

Wireless Power: a Path Toward Core-Less Transformers

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IEEE - PES
Presentation

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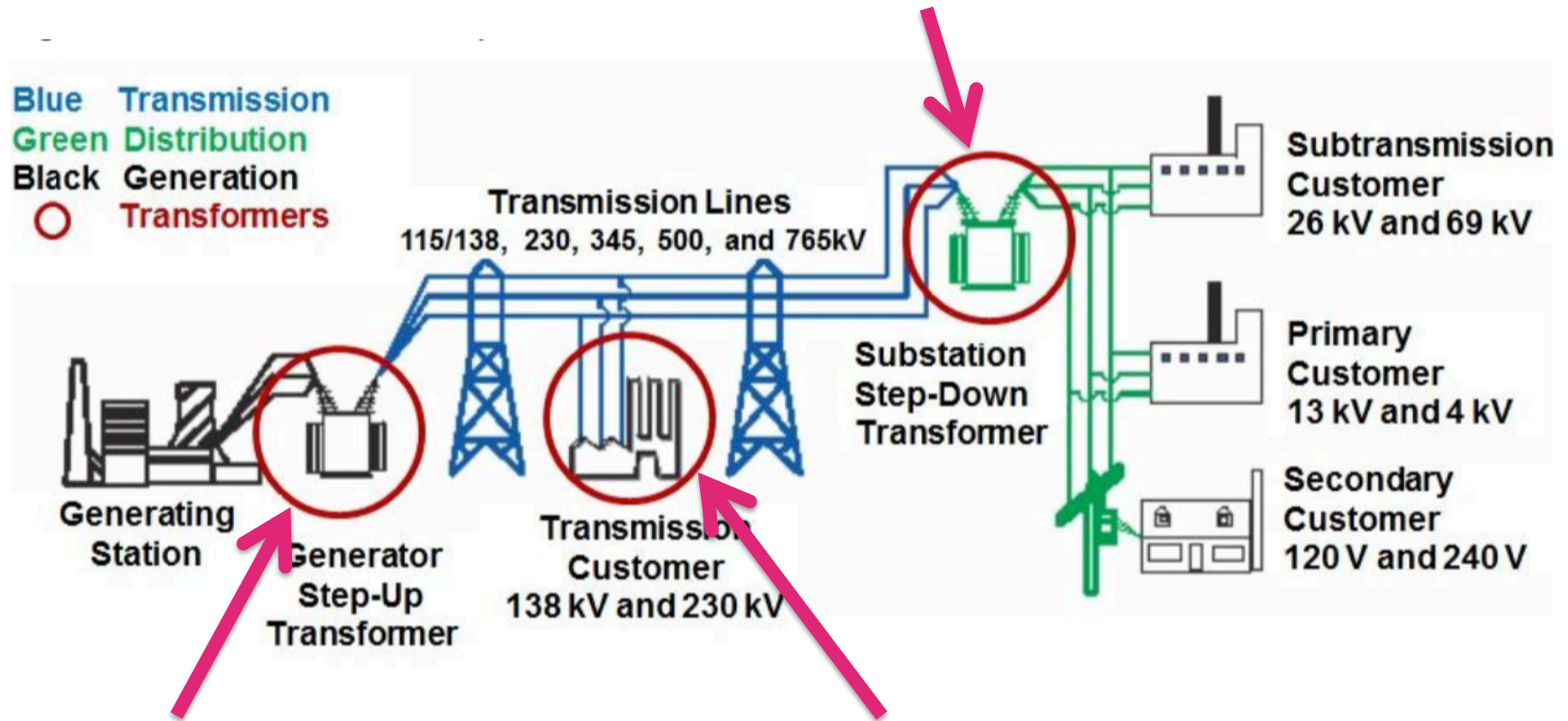
Cites: Feb2016



Presentation Outline

- Transformers:
 - Necessary Classical Device
 - Construction, Size, Core Basics
- An Alternate Approach:
 - Wireless Power: “Resonance”
- Resonant Coupling
- Scaling:
 - Resonant Coupling to High Power
- Key Limits, Future Prospects

Transformers: A Necessary Component In Electric Utility Systems



via: DOE 2012

“U.S. DOE report on large power transformers and the U.S. electric grid”

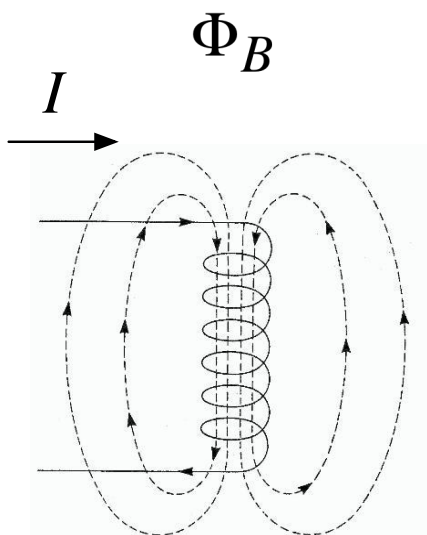
Transformers: A Classical Magnetic Device

- Transformer Basics
 - Magnetics: The coupling of 2 windings
 - Use of steel core, tight magnetic coupling
- Structure and Internals
- Costs, Weight (especially of steel core)
 - Worldwide manufacture and shipping
- Size and Scaling Relations
 - Steel core

Transformers: Linked Magnetic Flux

$$V_2 = -N_2 \frac{d\Phi_{B2}}{dt} \quad \text{Faraday's Law}$$

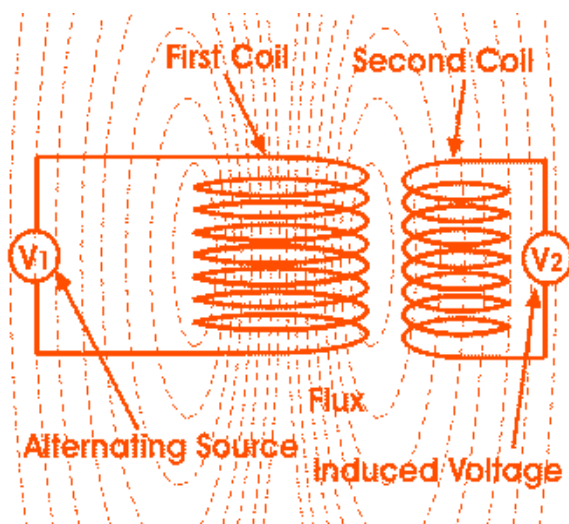
Magnetic Flux



Single
Coil

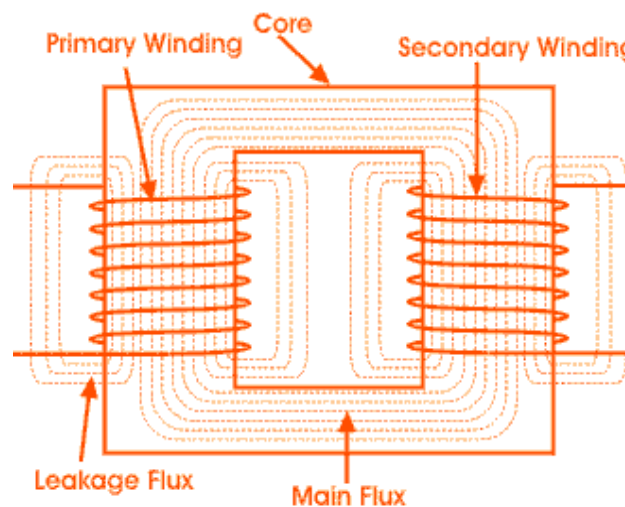
Φ_{B1}

Φ_{B2}



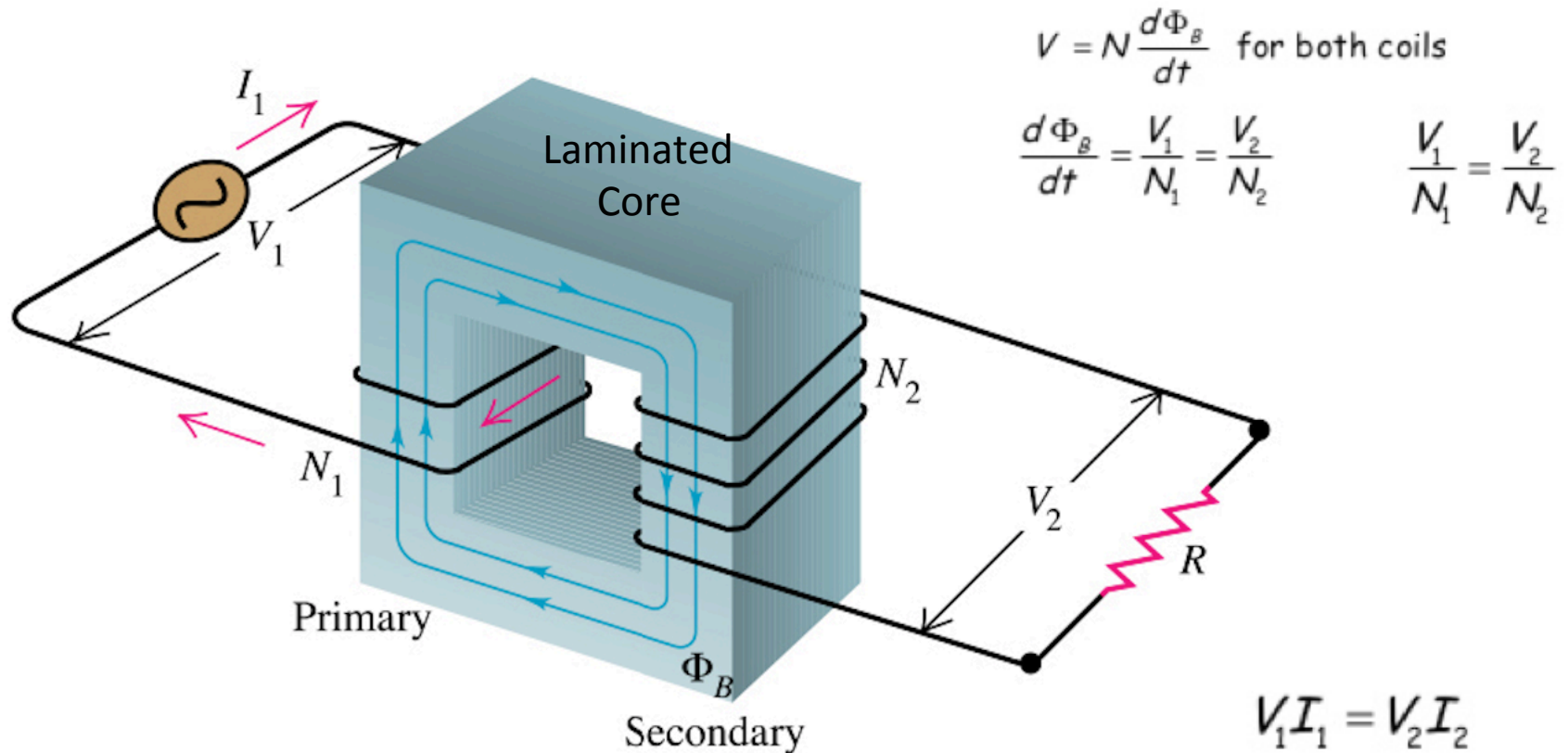
Coupled
Coils

$\Phi_{B2} \cong \Phi_{B1}$



Tightly Coupled
Coils

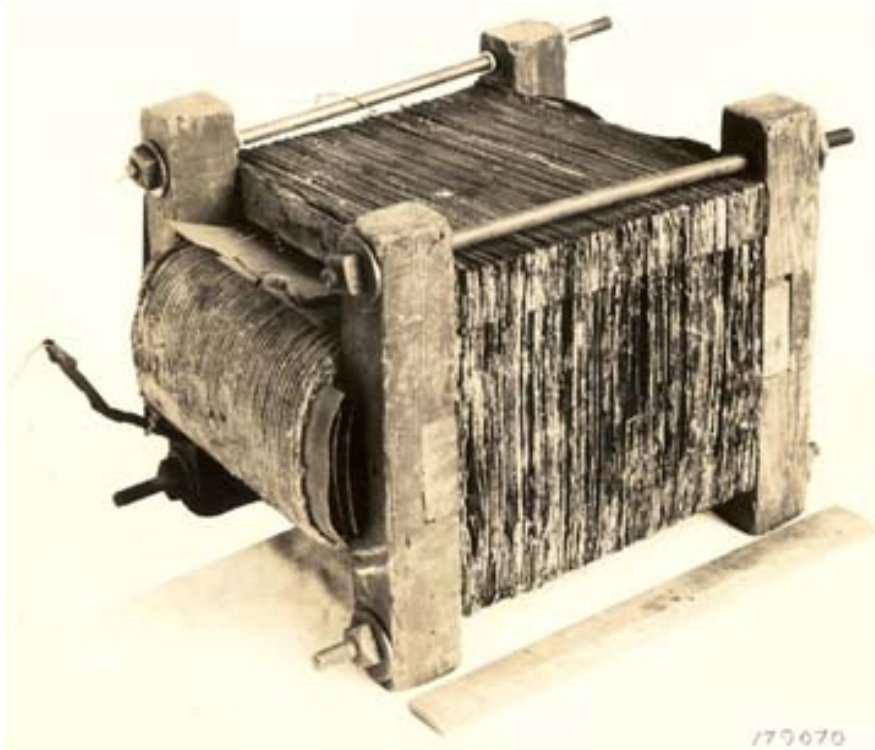
Basic Transformer Structure



via: www.indiastudychannel.com

In The Beginning

Stanley: 1886 'E'-core transformer



BDZ: 1885 toroid-core transformer

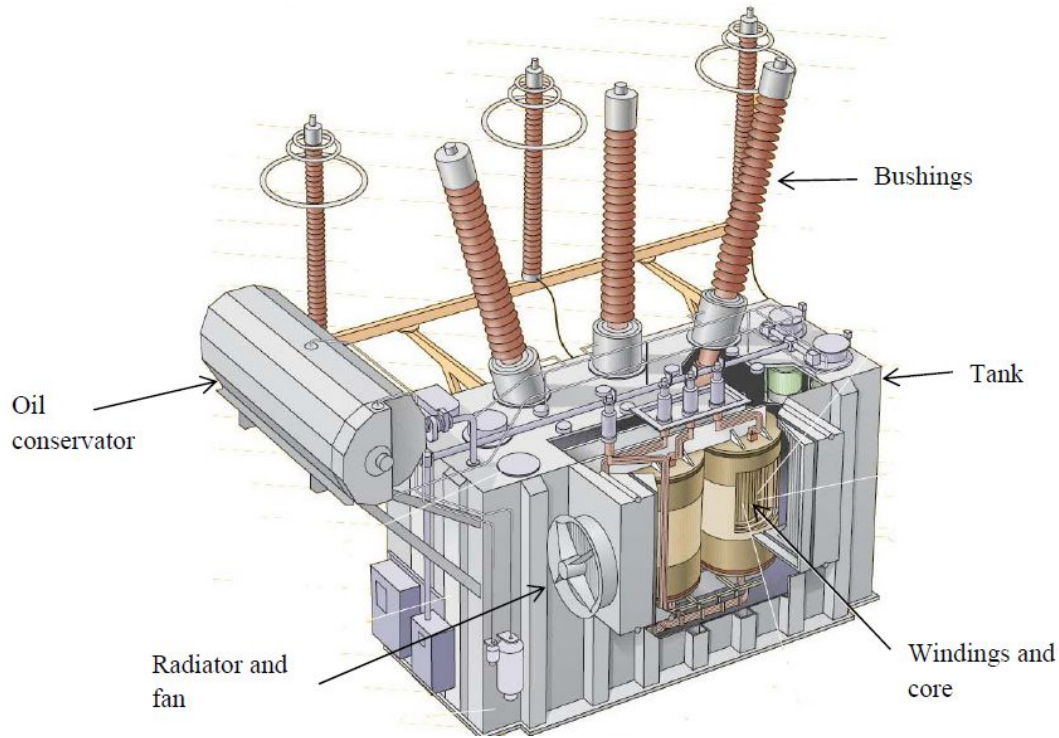


Typical Transformer Coils On Core



<http://www.transformerindia.com/photos/cb04.jpg>

Power Transformer Structure



112 MVA

via: www.cbsa-asfc.gc.ca

via: www.meppi.com/Products/Transformers/



Transformer Steel Cores



Steel Sheet Rolls

<http://siliconsteel.com/>



Cut and Stacked

via: www.tehenergo.ee/en/item/36-distribution_transformers.html

Steel Core Assemblies



via: www.lagor.it

Steel Core Assemblies

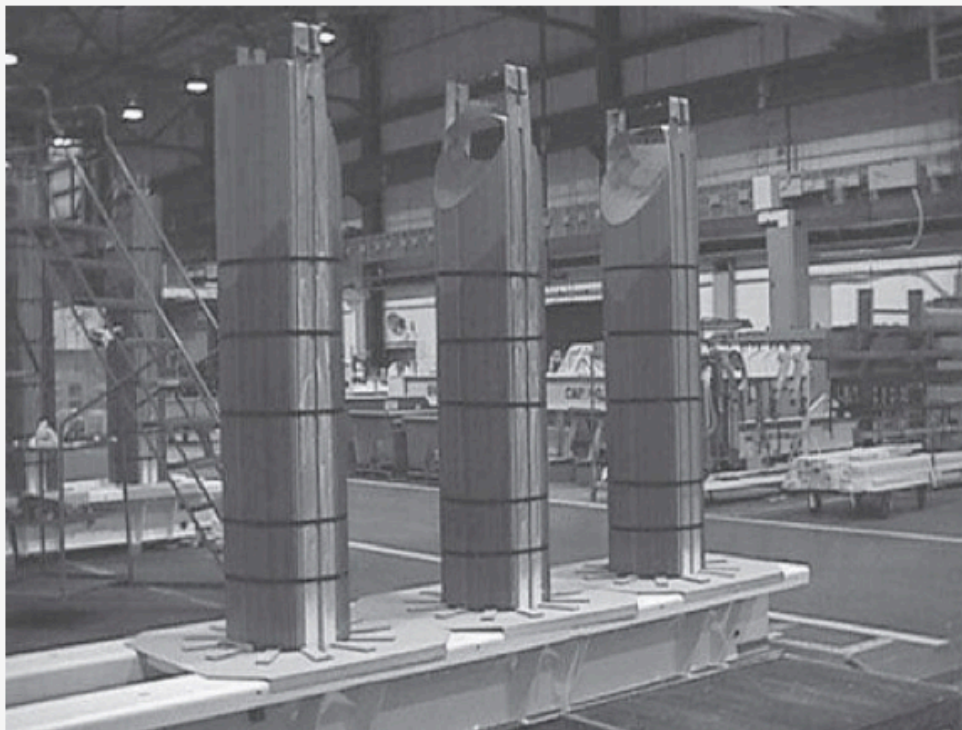


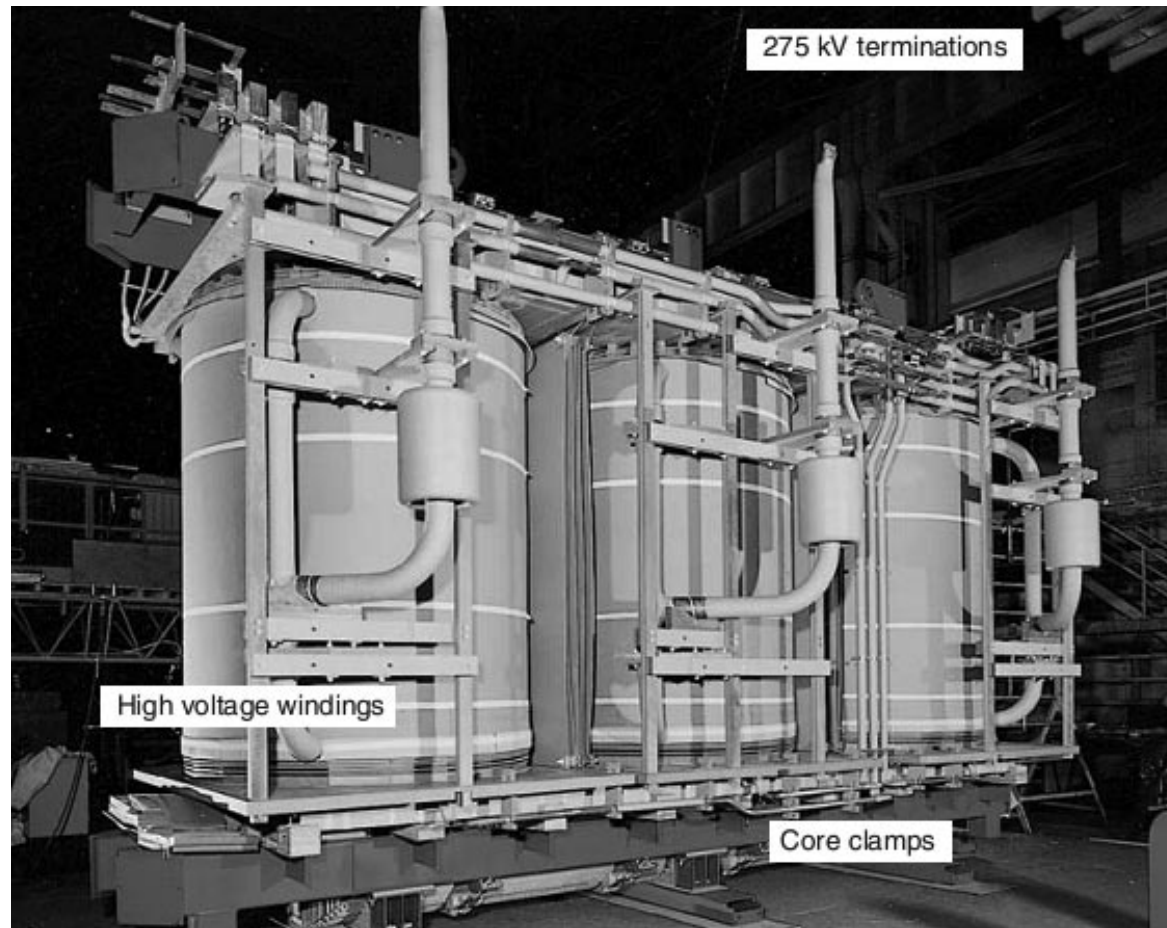
Figure 3 – 'E'-assembly, prior to addition of coils and insertion of top yoke

via: electrical-engineering-portal.com/power-transformer-construction-core



via: www.copper.org/

Full Construction Over Steel Core



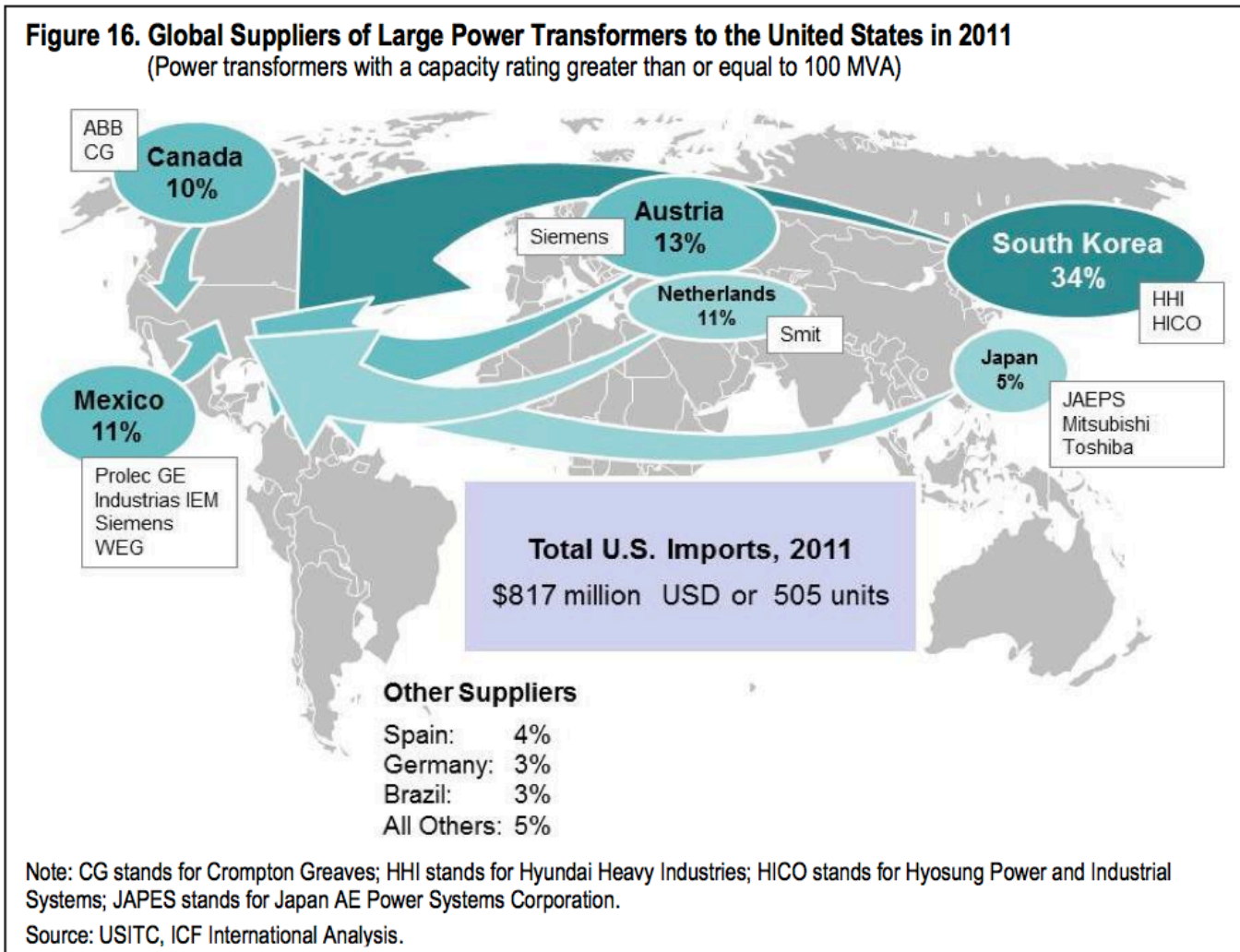
via: electrical-engineering-portal.com/what-are-the-main-classes-of-transformer

Transformer Costs: 100-500 MVA

- Total Cost: 2 to 5 \$million [~ \$10/lb]
- Total Weight: 200,000 to 600,000 lbs
- Cost Breakout:
 - Raw Materials: ~60%
 - Copper: ~25%
 - Core Steel: ~25%
- Weight Breakout
 - Core Steel: ~200,000 lb (larger MVA)

via: 2011 Data, DOE 2012

Transformers Travel to US



Transformer Size Relationships

- Basic Magnetics Theory: Faraday and Ampere
- Two Specific 'Areas' related to Power Capacity
- Power vs Size: Scaling relations
- Frequency effects
- Permeability effects (linear like frequency)
- Cooling more severe at higher powers

Transformer Equations

Voltage:

$$E = \frac{d\Phi}{dt} = \frac{1}{\sqrt{2}} \omega N B A_C = \frac{2\pi}{\sqrt{2}} N f B A_C \quad (\text{volts})$$

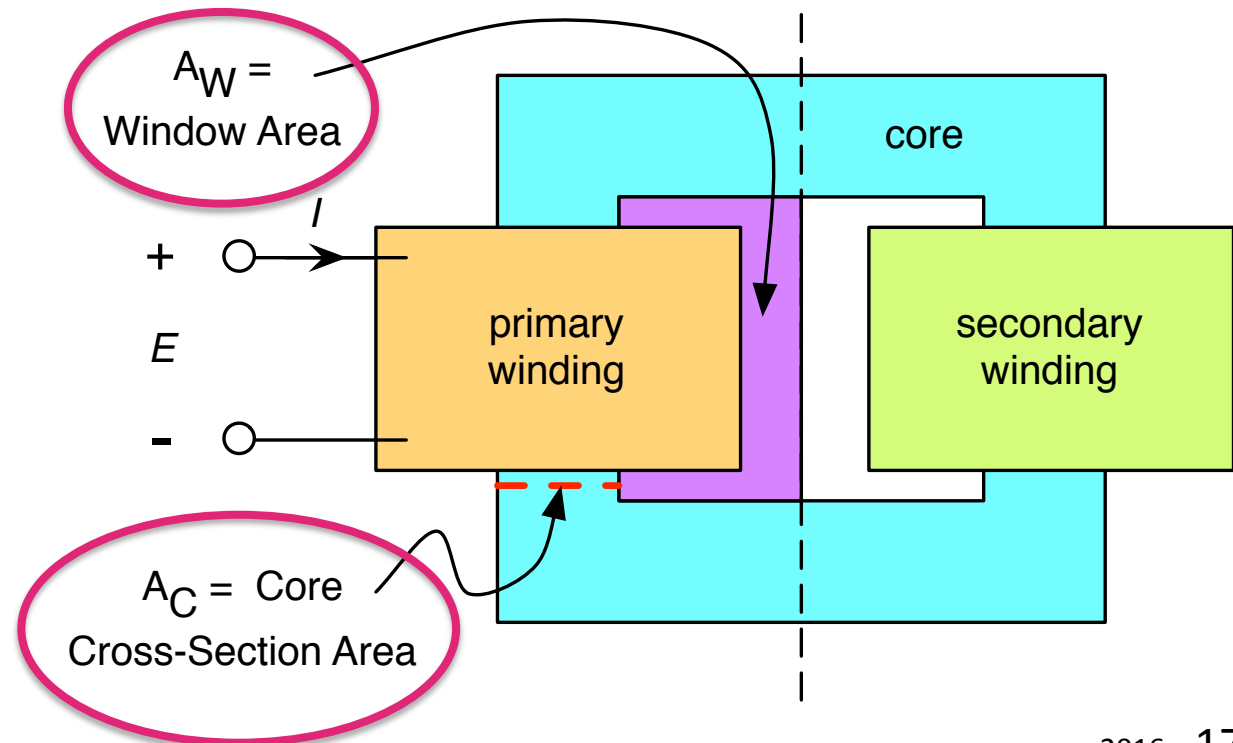
Current:

$$I = J_{\text{wire}} A_W \quad (\text{amp})$$

Core Area

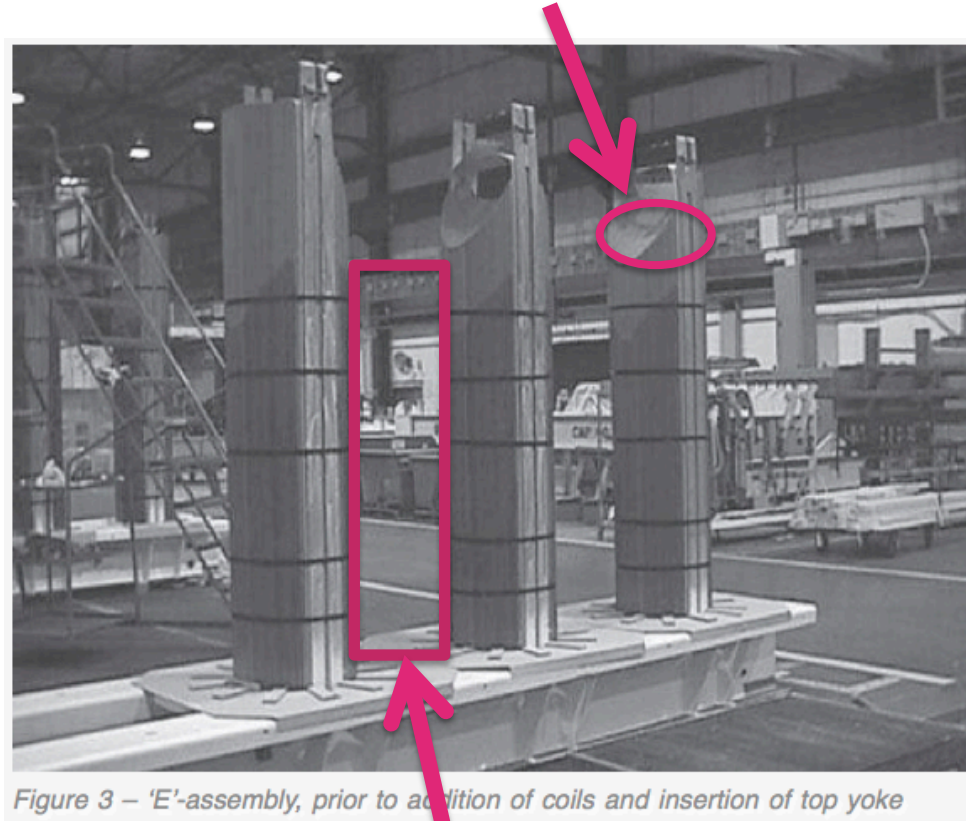
Flux Density

$$P = E I$$



Key Specific 'Areas' for Power Capability

Core Cross-Section Area, A_c



'Winding Window' Area, A_w

via: electrical-engineering-portal.com

Transformer Power, P (watts)

Window Area Core Area

$$P = EI = KfB J_{wire} A_W A_C \quad (\text{watts})$$

But: $A_W = K_U A_C$

with:

$B_{\max} = 1.1$ Tesla, $J_{\text{wire}} = 500$ Amp/cm², 40% winding fill factor

f = frequency

A_C = Core Area

Power:

$$P = 0.56 f (A_C)^2 \quad (\text{watts})$$

f in hertz,
 A_C in in²

Examples: Transformer Size - Power

$$P = 0.56 f (A_C)^2 \quad (\text{watts})$$

At 60 Hz:

$$\text{Core Area} = 100 \text{ in}^2 \quad \rightarrow \rightarrow \quad 336 \text{ kW}$$

$$\text{Core Area} = 1000 \text{ in}^2 \quad \rightarrow \rightarrow \quad 33.6 \text{ MW}$$

$$\text{Core Area} = 3000 \text{ in}^2 \quad \rightarrow \rightarrow \quad 302 \text{ MW}$$

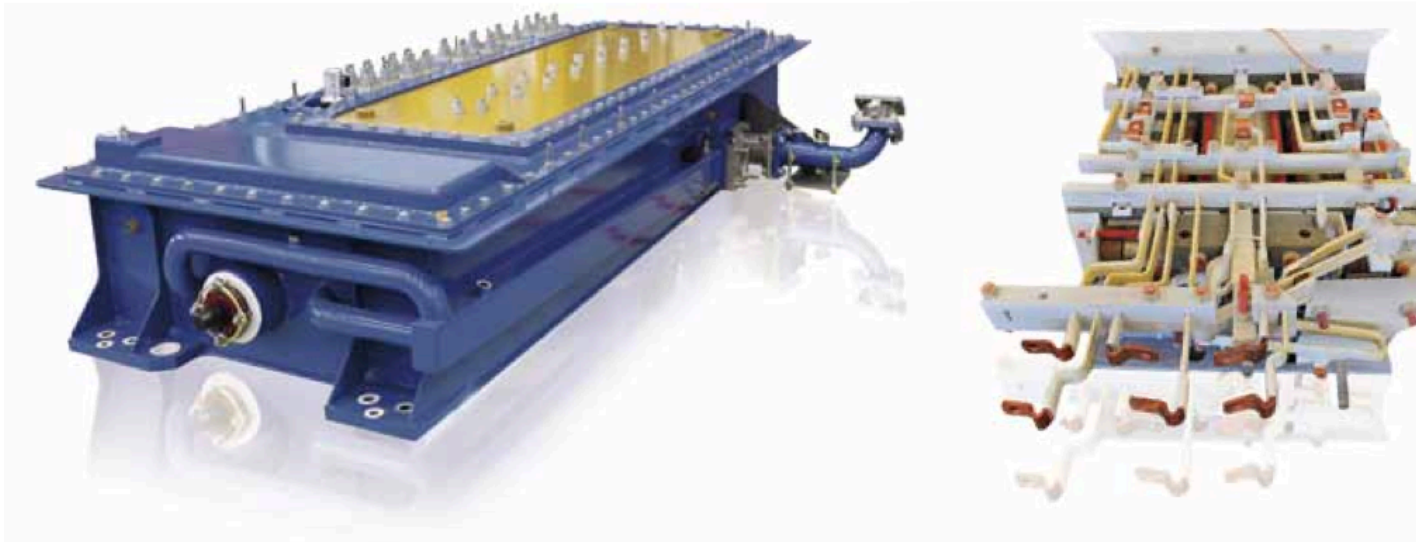
At 6 kHz:

$$\text{Core Area} = 100 \text{ in}^2 \quad \rightarrow \rightarrow \quad 33.6 \text{ MW}$$

Higher Frequency \rightarrow Smaller Size

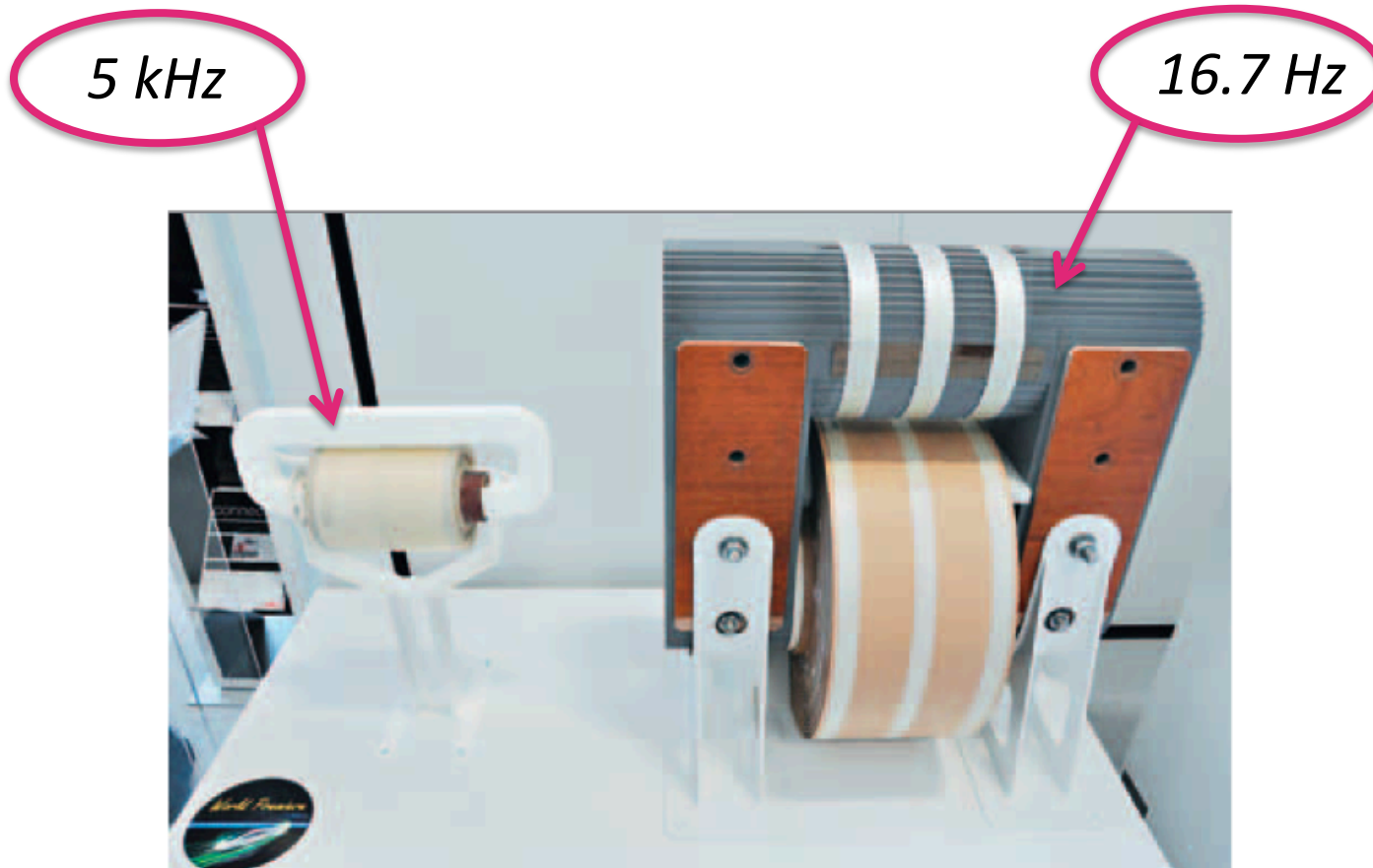
- Recall: $P = 0.56 f (A_C)^2$ (watts)
- Examples for Traction Transformers

450 – 675 kW: at 1.75 kHz
Via: ABB



via: 'SR Transformers-121031.pdf'

Size – Frequency Relations



Left: 5 kHz Transformer. Right: 16.7 Hz Transformer. Both: 3.6 kV Primary, 1.5 kV Secondary.

via: ABB Brochure: "Shrinking the Core"

An Alternate Approach to Transformers

- Recall: Classical Transformer Approach
 - ‘Unity’ magnetic coupling via steel core
 - Steel core about 25% of weight and cost
 - Core-Losses = major source of heat
- New Approach – Technology from Wireless Power Systems
 - No physical contact between ‘primary’ and ‘secondary’
 - No core
 - Not ‘unity’ magnetic coupling
 - Higher frequency operation
- Achieved via ‘Resonant Coupling’
 - Originator = Tesla

Resonant Power - The 'Pioneer' = N. Tesla

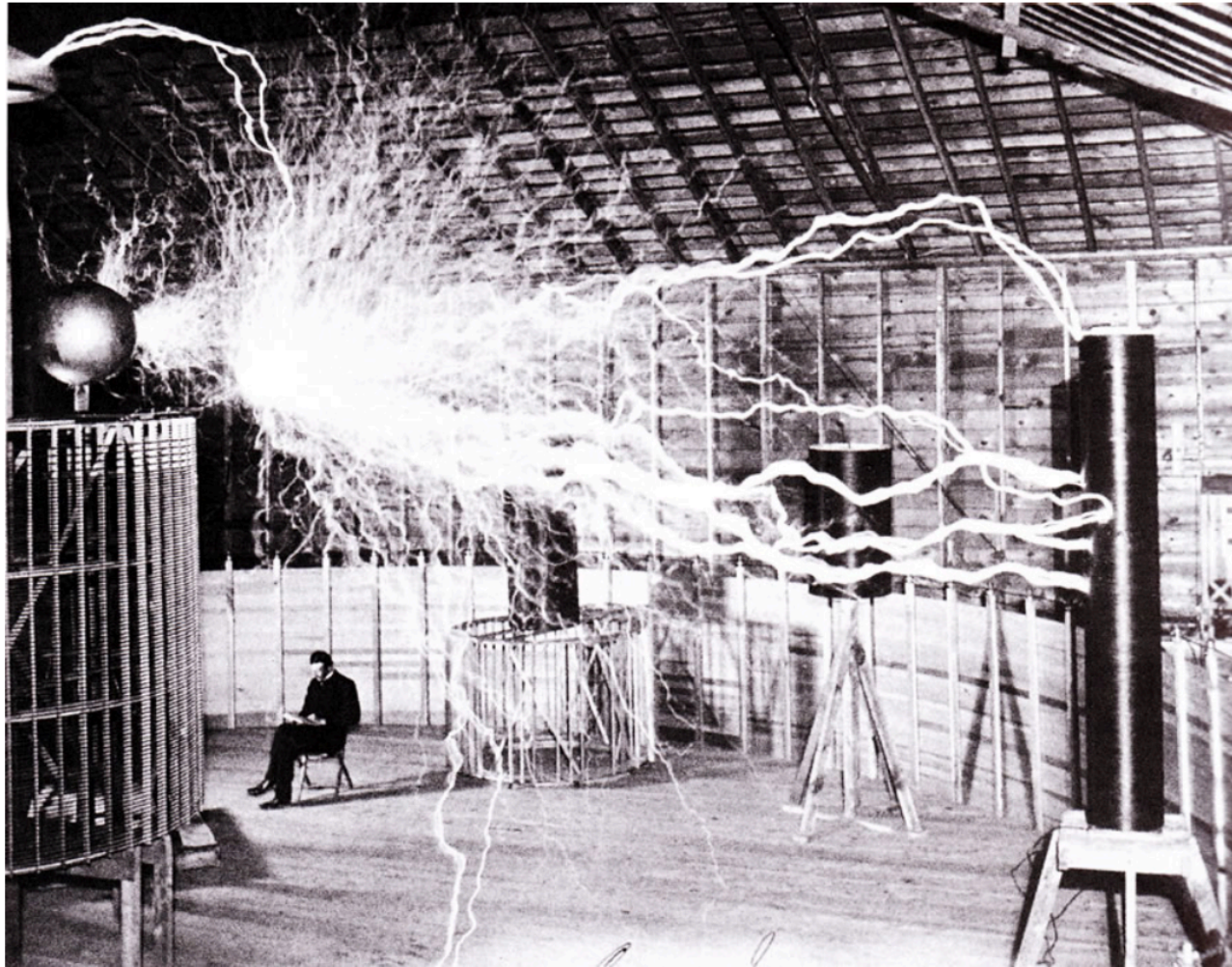
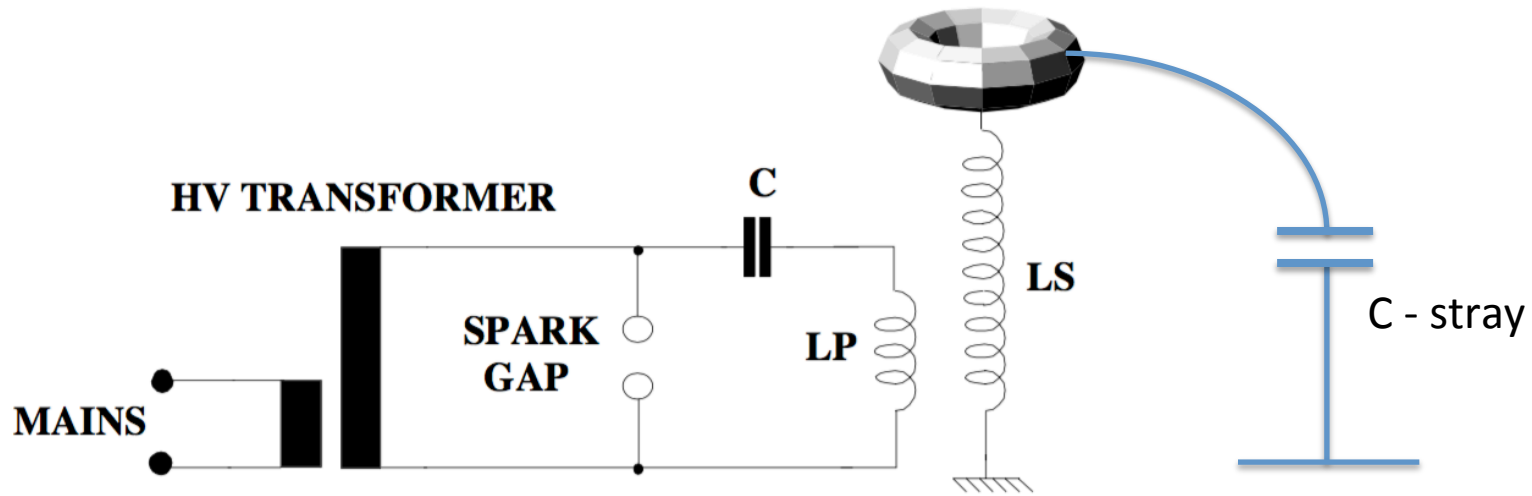
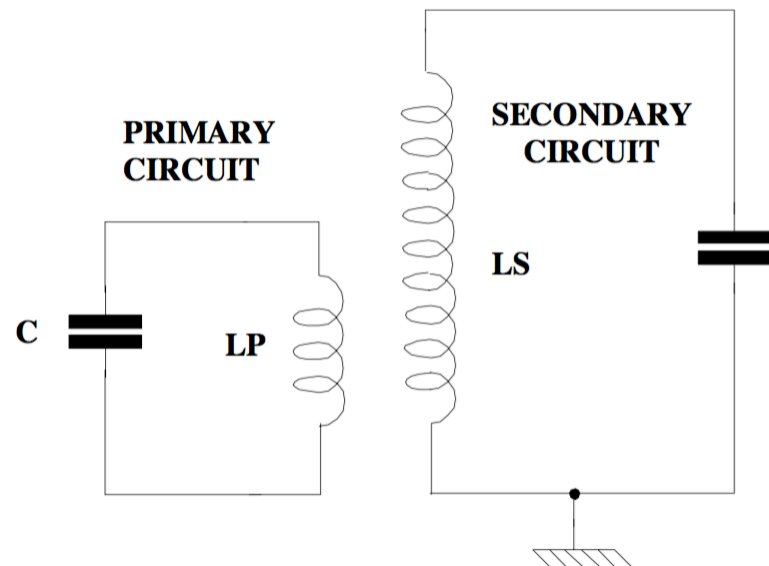


Figure 2.2: Tesla's Colorado Springs experiments (from [26])

Tesla: Double Resonant Coils

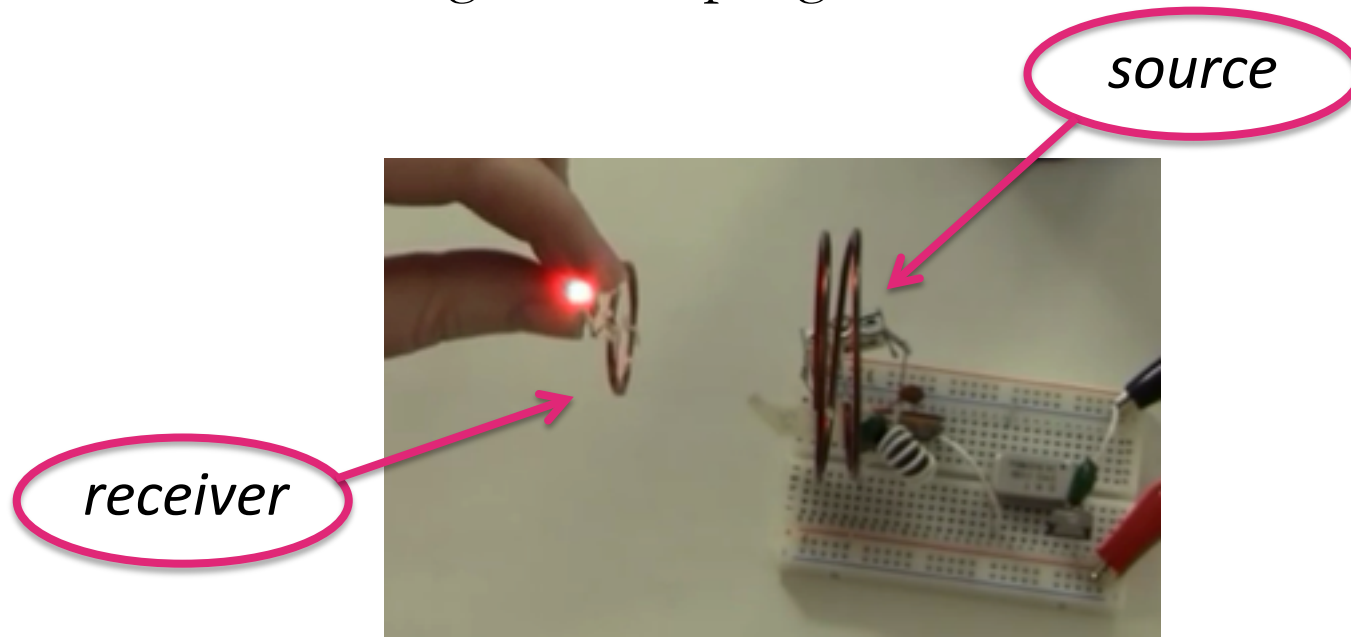


Two Magnetically Coupled Resonant L-C Circuits



Key Concepts for Resonant Power

- Use 1st resonant L-C circuit as source of magnetic energy
- Use 2nd resonant L-C to collect magnetic energy
- Need 'weak' magnetic coupling between the two L-C circuits



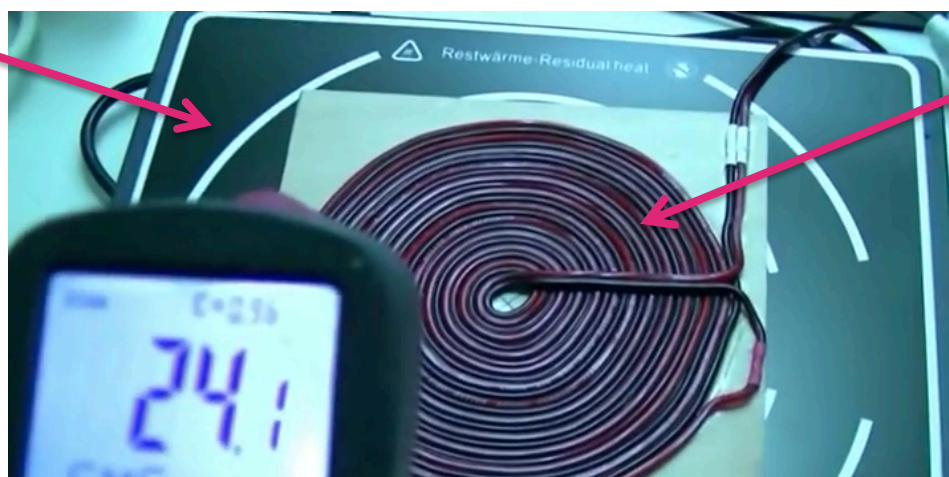
via: www.youtube.com/watch?v=r1UT4NuygmQ

Induction Cooktop Yields 4kW Wireless Power



*Induction
Cooktop
= Source*

*Receiver
Coil*



Wireless Power Examples: Buses, Cars

United Kingdom – Wireless Milton Keynes electric bus route.
January 2014



Source: <http://transportevolved.com/2014/01/09/milton-keynes-uk-launches-wirelessly-charged-electric-bus-route/>

| | |
|-----------------------|--------|
| Frequency | ?? Hz |
| Power Transfer Levels | 120 kW |
| Wireless Mode | static |

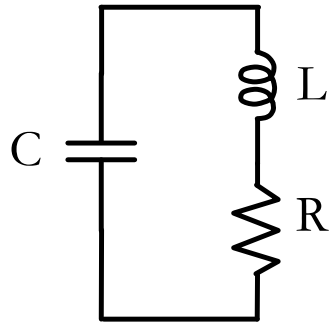
Belgium – Volvo & Bombardier test inductive charging.
October 2013



Source: <http://evworld.com/news.cfm?newsid=31633>

| | |
|-----------------------|---------|
| Frequency | 140 kHz |
| Power Transfer Levels | 20 kW |
| Wireless Mode | static |

Resonance Equations – Single L-C

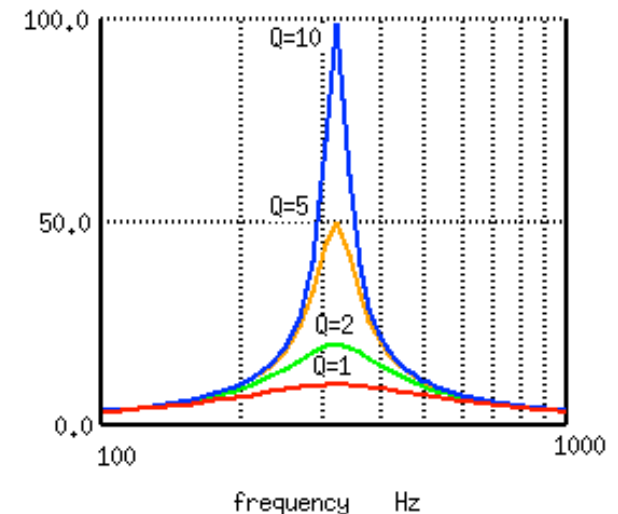


$$\omega_0 = \frac{1}{\sqrt{LC}}$$

$$H(s) = \frac{\omega_0^2}{s^2 + \frac{\omega_0}{Q}s + \omega_0^2}$$

$$Q \stackrel{\text{def}}{=} \frac{f_r}{\Delta f} = \frac{\omega_r}{\Delta\omega}$$

Q = Quality Factor



$$Q \stackrel{\text{def}}{=} 2\pi \times \frac{\text{Energy stored}}{\text{Energy dissipated per cycle}} = 2\pi f_r \times \frac{\text{Energy stored}}{\text{Power loss}}$$

$$Q(\omega) = \omega \times \frac{\text{Maximum energy stored}}{\text{Power loss}}$$

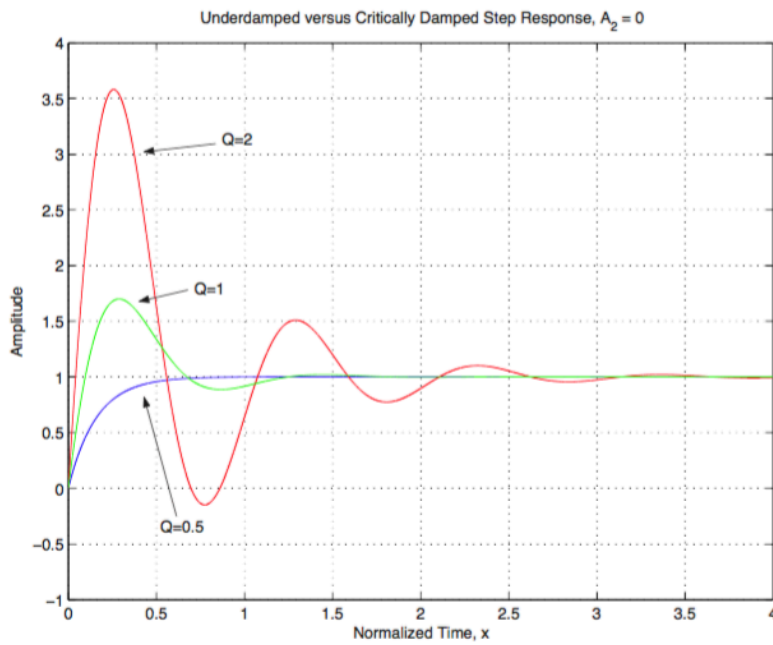
$$Q = \frac{X_L}{R}$$

Series R-L-C Circuit

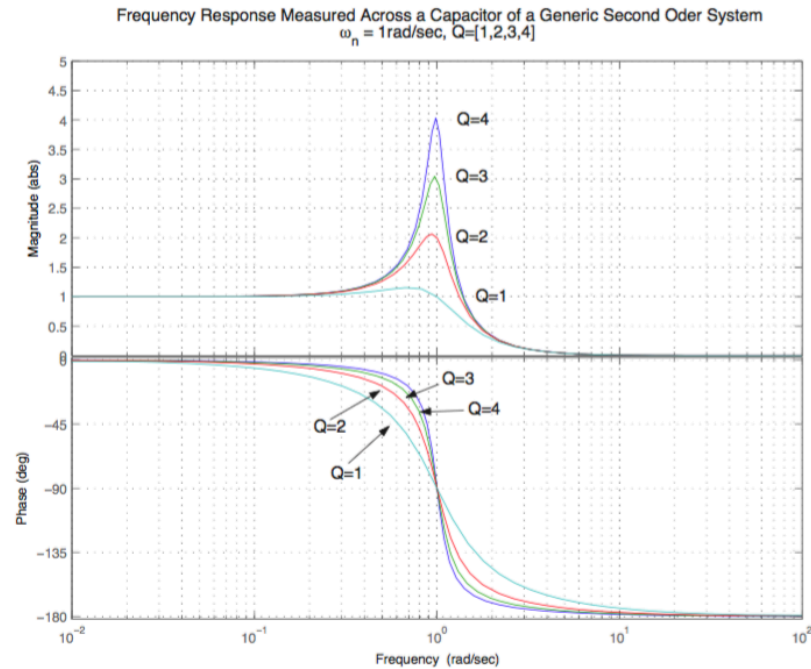
Single L-C: Time and Frequency Responses

$Q > 1$

Oscillatory Time Domain Response



Peaked Frequency Domain Response



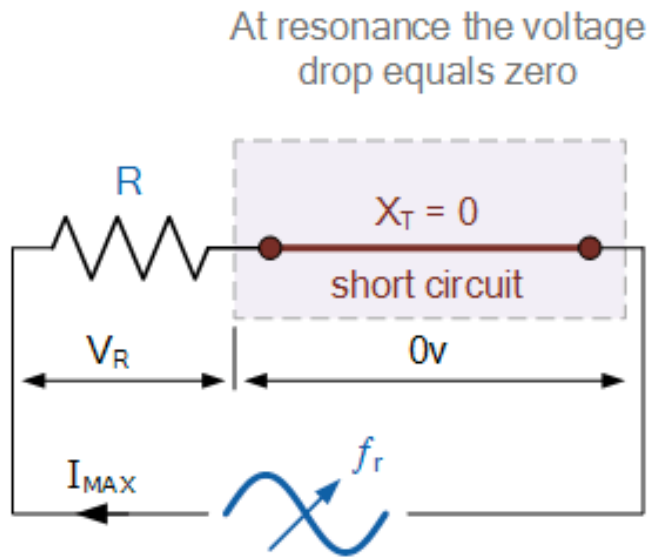
$$y(x) = e^{-\frac{\pi x}{Q}} \left(A_1 \cos\left(\frac{\pi x}{Q} \sqrt{4Q^2 - 1}\right) + A_2 \sin\left(\frac{\pi x}{Q} \sqrt{4Q^2 - 1}\right) \right)$$

$$\frac{V_{out}(s)}{V_{in}(s)} = \frac{\omega_n^2}{s^2 + \frac{\omega_n}{Q}s + \omega_n^2}$$



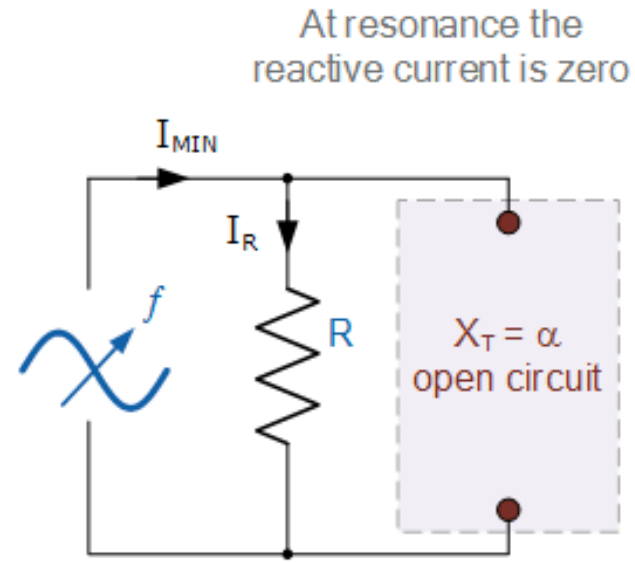
At Resonance; L & C Reactance 'Cancel'

Series = Short Circuit
(zero voltage)



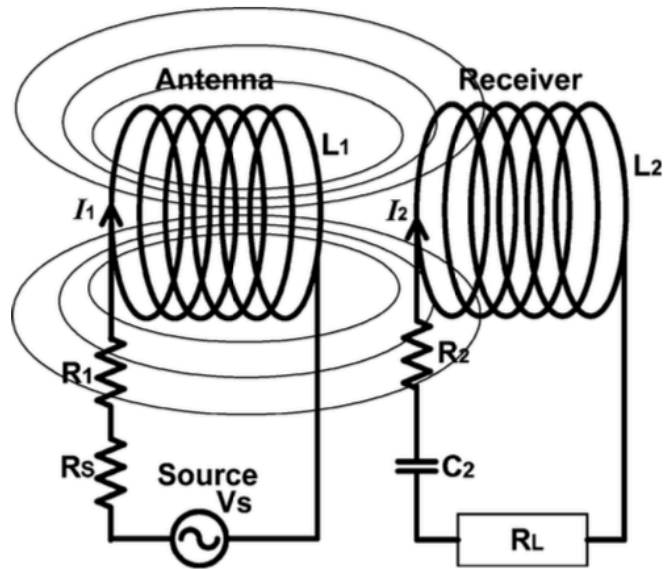
$$Q = \frac{X_L}{R}$$

Parallel = Open Circuit
(zero current)



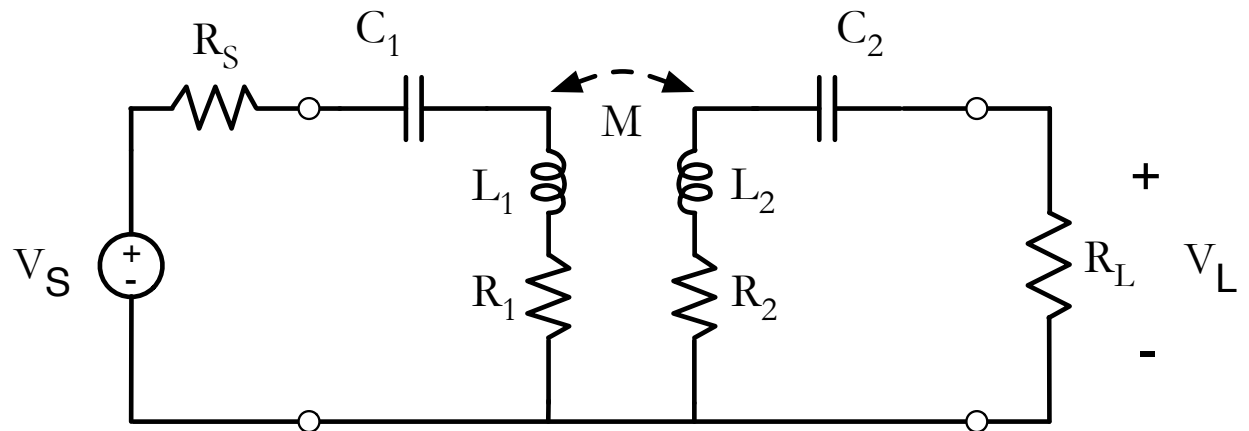
$$Q = \frac{R}{X_L}$$

Double Coupled L-Cs Resonant Circuits



Weakly linked magnetic flux

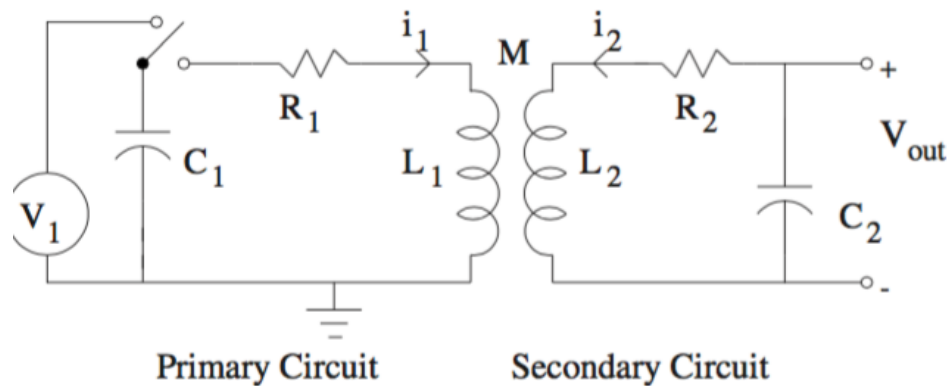
Coupled Equivalent Circuit:



Primary L-C

Secondary L-C

4th Order Equations – ‘Tesla Coil’

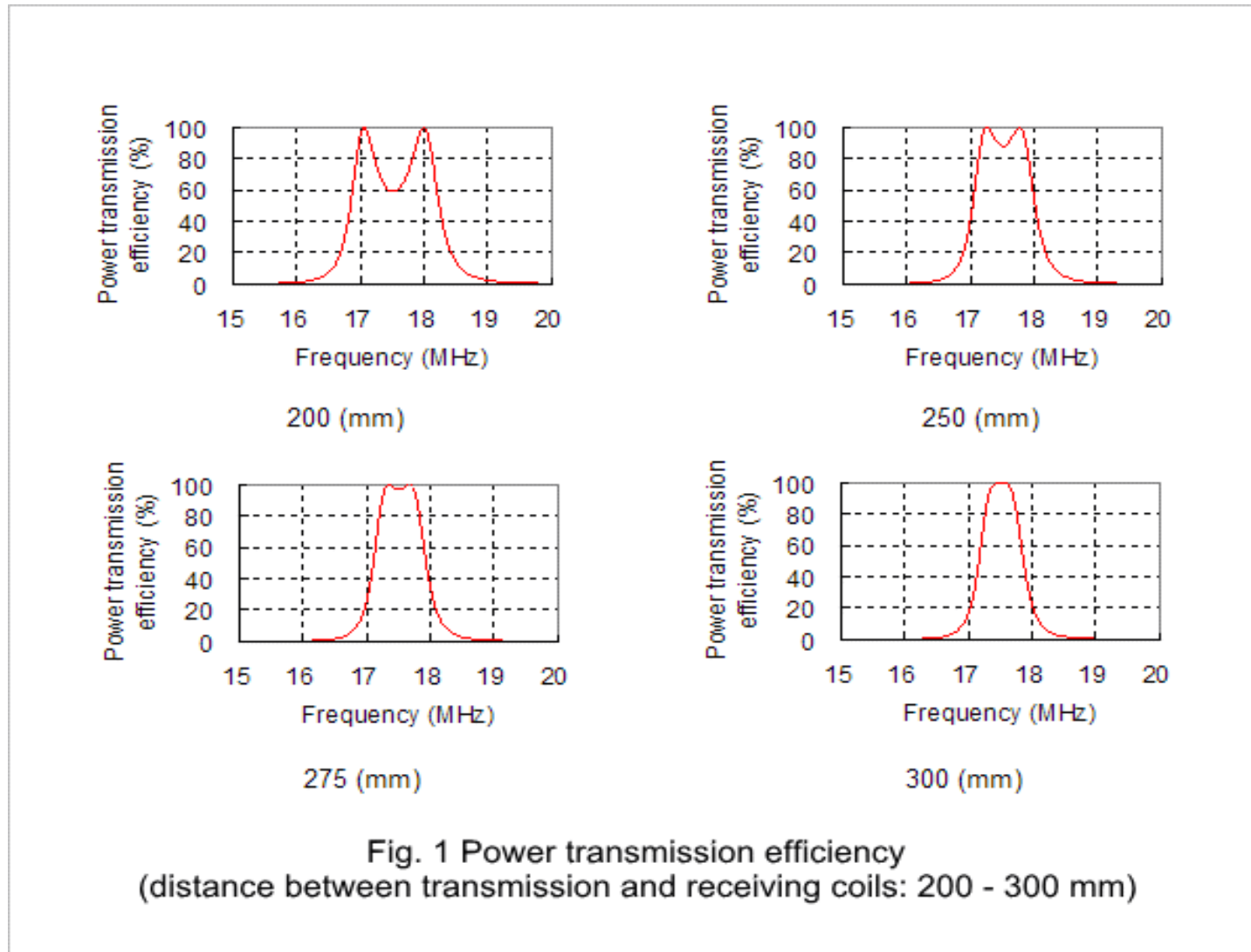


4 Energy Storage Elements

$$\frac{V_{out}}{V_{in}} = -\sqrt{\frac{L_2}{L_1}} \frac{s^2 \omega_2^2 k}{\left\{ \begin{array}{l} (1 - k^2)s^4 + \left(\frac{\omega_1}{Q_1} + \frac{\omega_2}{Q_2}\right)s^3 + \left(\omega_2^2 + \omega_1^2 + \frac{\omega_1\omega_2}{Q_1Q_2}\right)s^2 \\ + \omega_1\omega_2\left(\frac{\omega_2}{Q_1} + \frac{\omega_1}{Q_2}\right)s + \omega_1^2\omega_2^2 \end{array} \right.}$$

$$\omega_i = \frac{1}{\sqrt{L_i C_i}} \quad \text{and} \quad Q_i = \frac{\omega_i L_i}{R_i} \quad M = k\sqrt{L_1 L_2}$$

Response: Single or Double Frequency Peaks



At Resonance: Highly Coupled

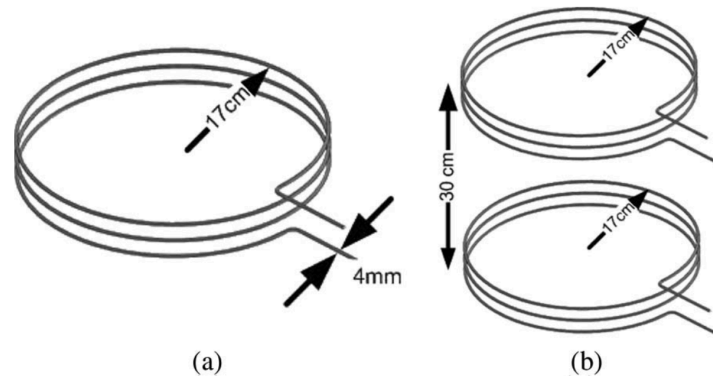


Fig. 6. Drawings of an isolated coil and mutually coupled coils. (a) Isolated coil. (b) Two coupled coils.

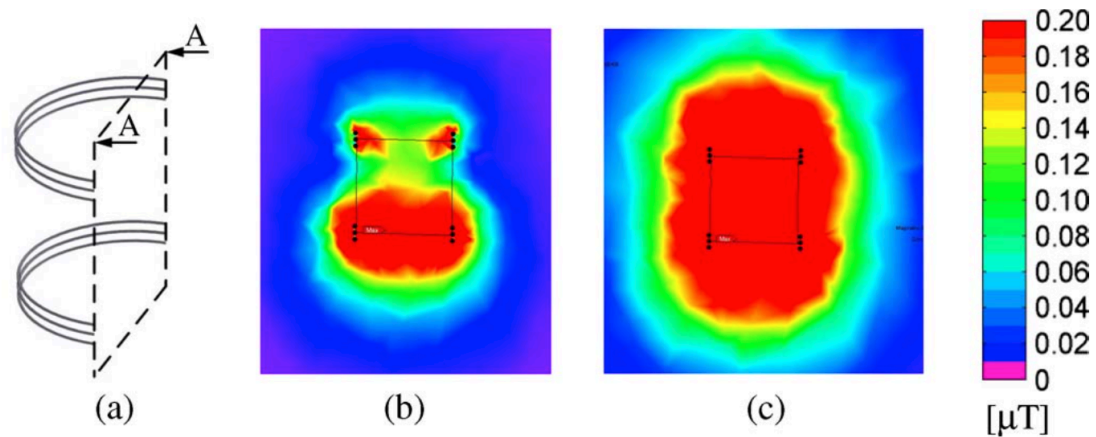
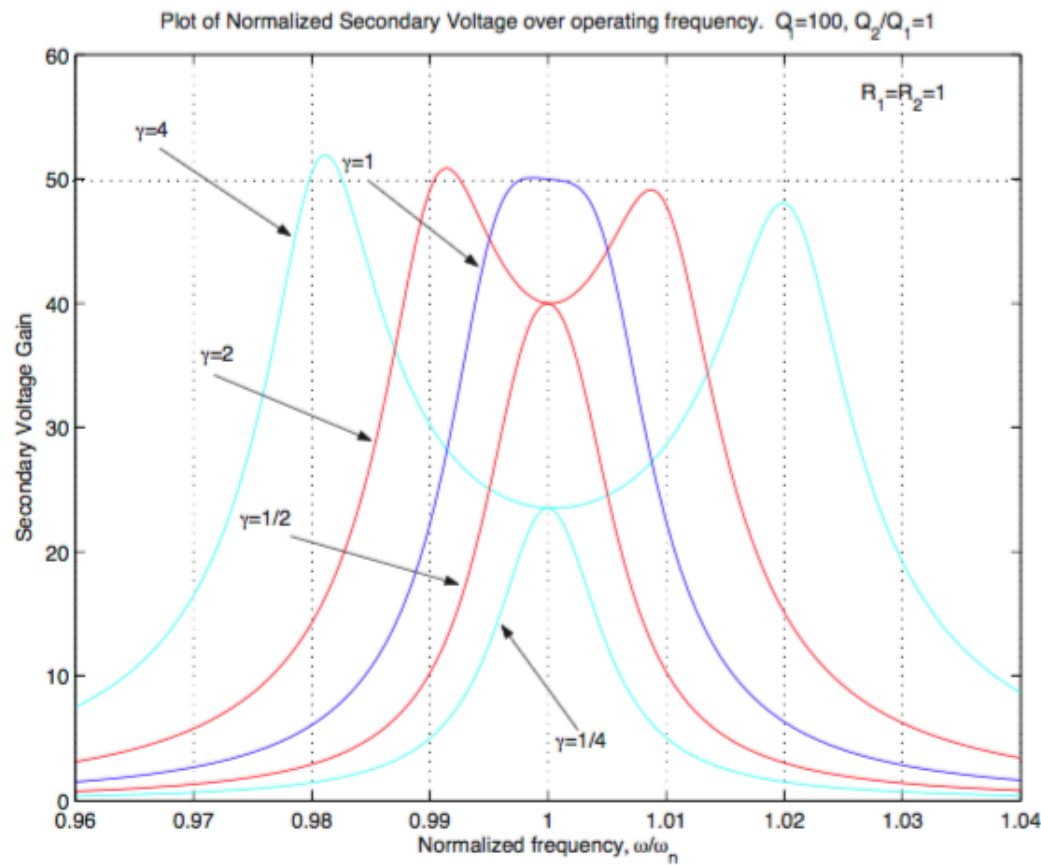


Fig. 7. Magnetic flux density dependence on operating frequency. (a) Sectional view. (b) 3.34 MHz. (c) 3.68 MHz (resonant frequency).

Coupling Changes Frequency Response



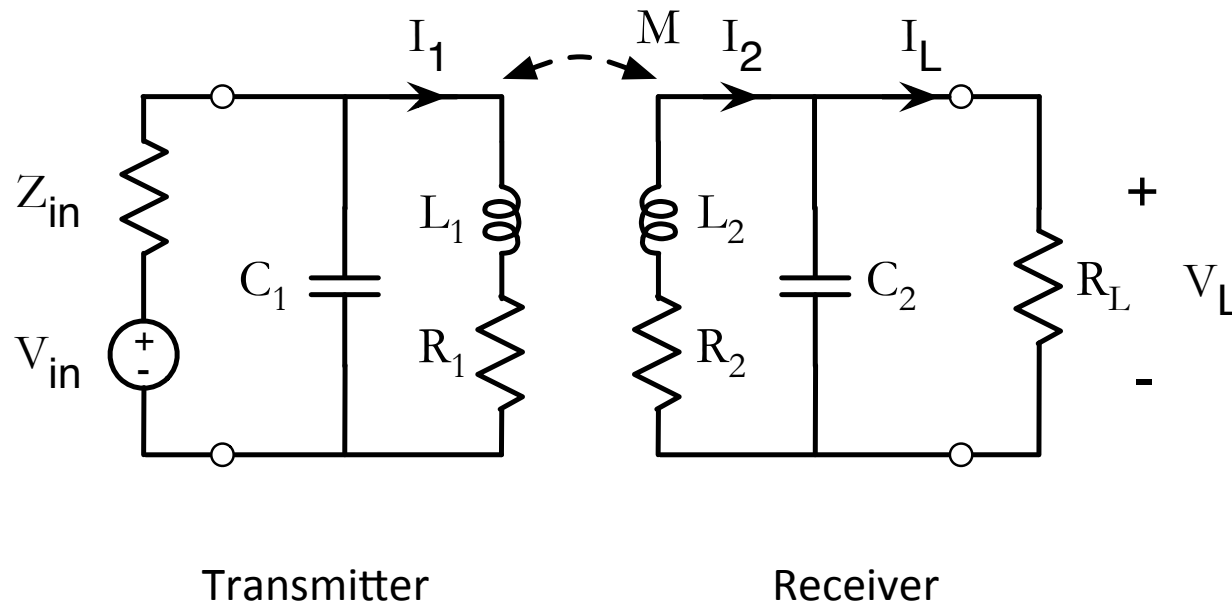
$$k \equiv \frac{M}{\sqrt{L_1 L_2}}$$

$$k_{crit} = \frac{1}{\sqrt{Q_1 Q_2}}$$

$$\gamma = \frac{k}{k_{crit}}$$



Power Efficiency: Double Resonant Circuit



$$\eta_1 = \frac{\text{Power to Load}}{\text{Total Power (Loss+Load)}} = \frac{I_L^2 R_{Le}}{I_1^2 R_1 + I_2^2 R_2 + I_L^2 R_{Le}}$$

Max Power Efficiency at Resonance

Calculate:

$$\eta_{1-\max} = \frac{k^2 Q_1 Q_2}{\left[1 + \sqrt{1 + (k^2 Q_1 Q_2)}\right]^2}$$

Figure-of-Merit

$$\text{FOM} = k \sqrt{Q_1 Q_2} \gg 1$$

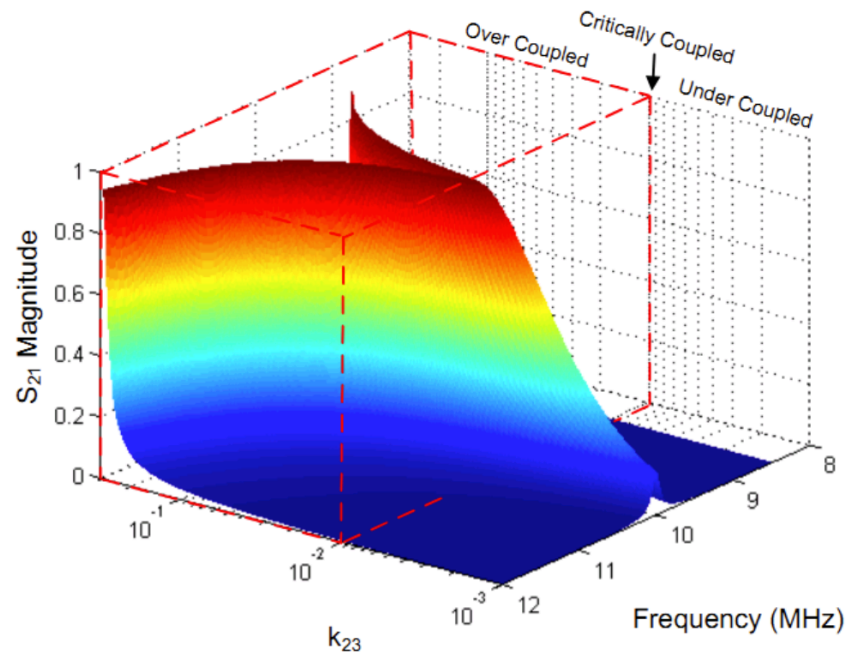
Where:

$$k \equiv \frac{M}{\sqrt{L_1 L_2}} \quad Q_1 = \frac{\omega L_1}{R_1} \quad Q_2 = \frac{\omega L_2}{R_2} \quad \omega = \begin{array}{l} \text{resonant} \\ \text{angular frequency} \end{array}$$

When at Optimum load resistance, $R_{L-\text{equiv-opt}}$

$$R_{L-\text{equiv}} = R_2 \left(1 + k^2 Q_1 Q_2\right)^{1/2}$$

Double Resonant Power Transfer, Optimization

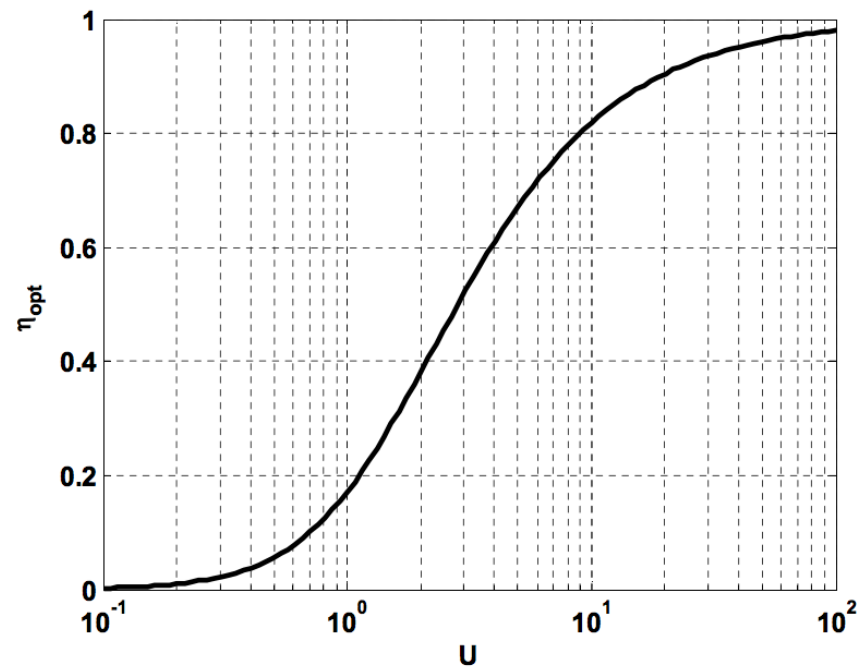


Sample: IEEE 2013

'Figure-of-Merit' = U

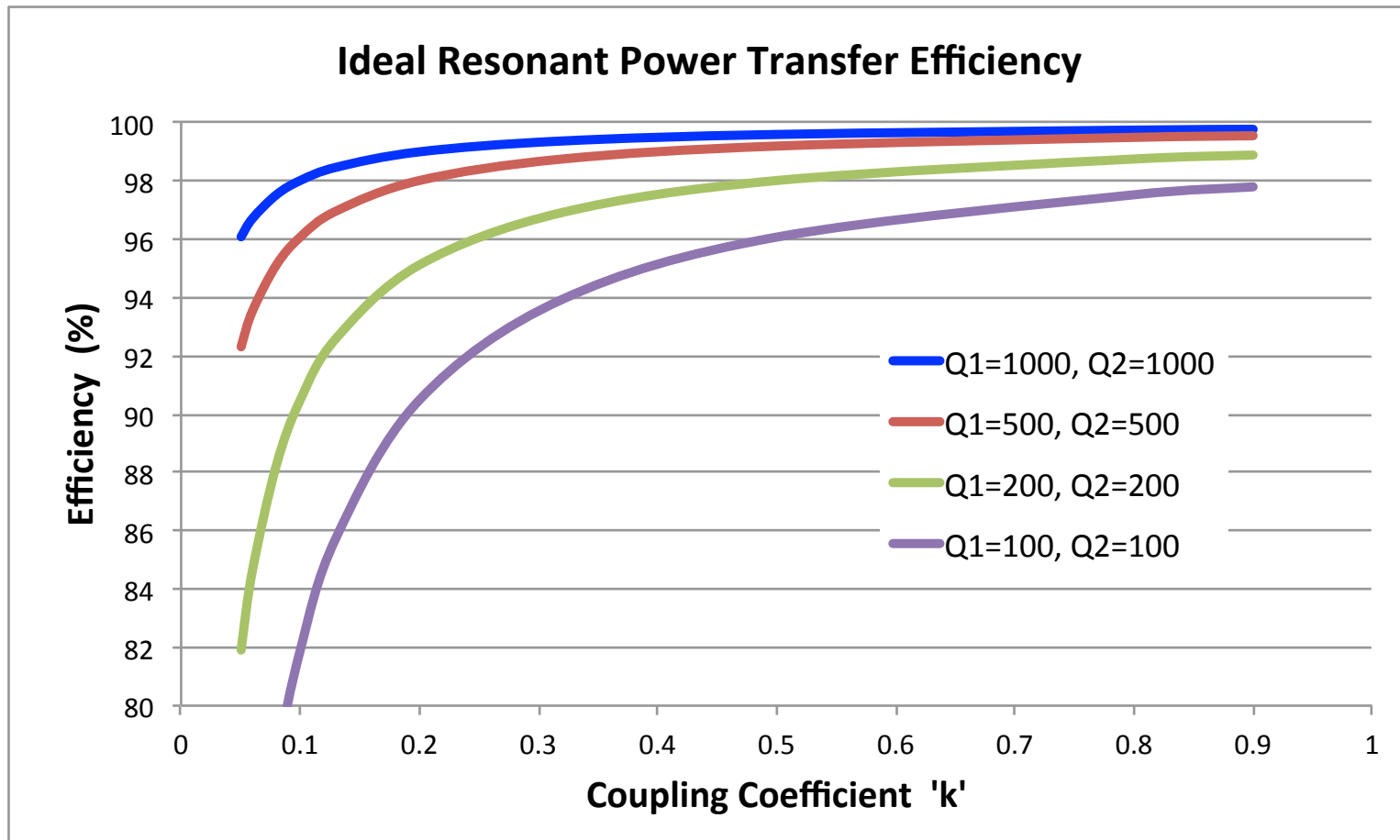
$$U = k\sqrt{Q_s Q_d}$$

$$\eta_{opt} = \frac{U^2}{\left(1 + \sqrt{1 + U^2}\right)^2}$$



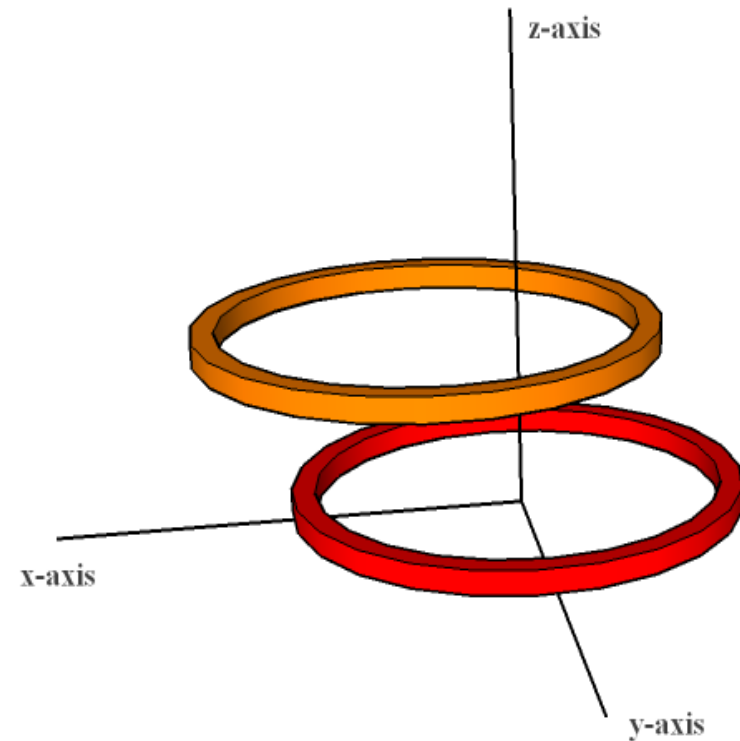
Kesler: Witricity 2013

Efficiency: 'k' Dependence with 'Q'

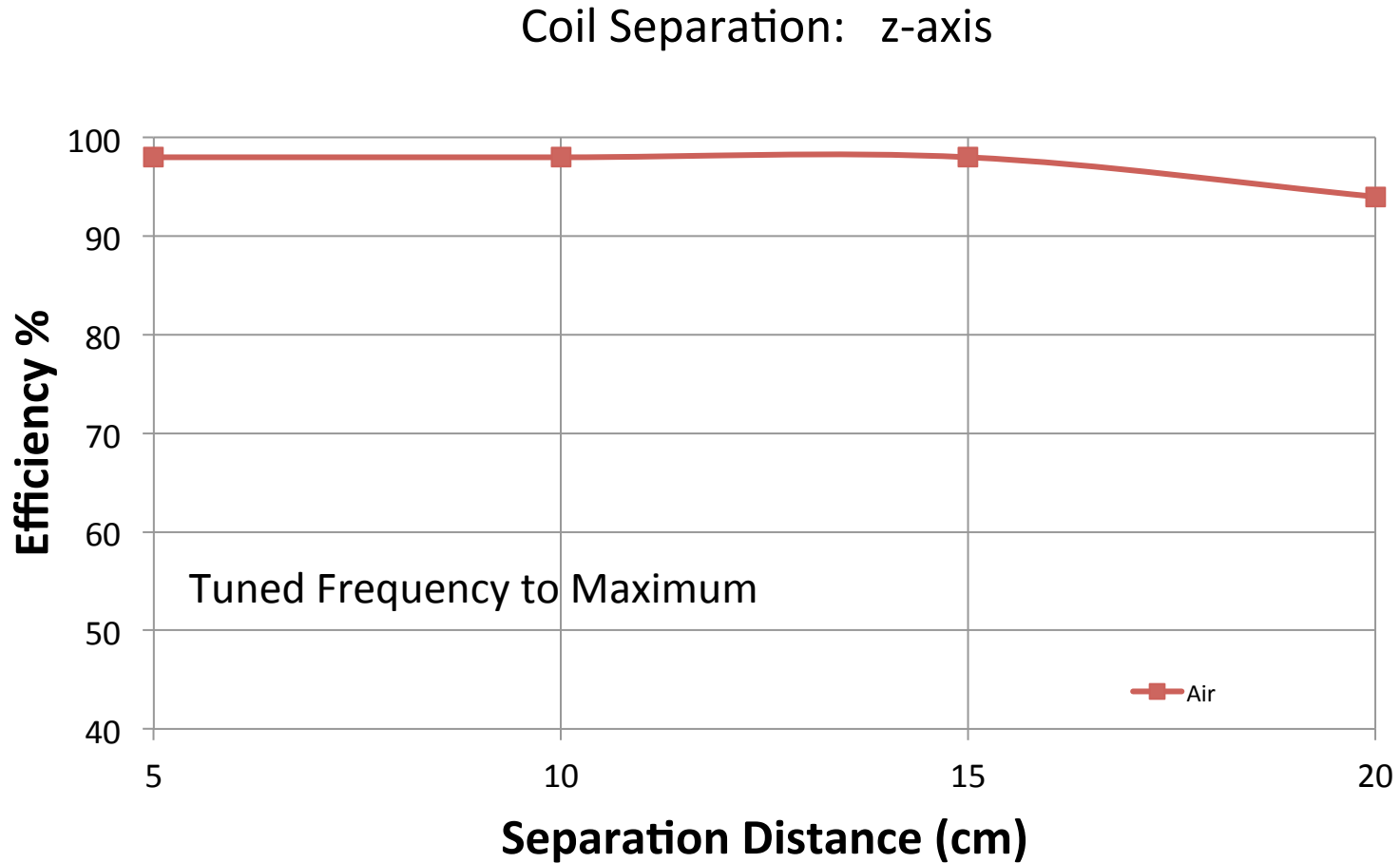


Coil Spatial Position Sensitivity Tests: Simple Circular Coils

- The transmitter (red) and receiver (orange) coils, aligned above one another and separated on the z-axis.
 - Separate: z-direction
 - ‘Shift’: x-direction
- Litz-wire coils are employed.
 - Symmetric, 30cm Diam. Coils.



Position Sensitivity: Separation



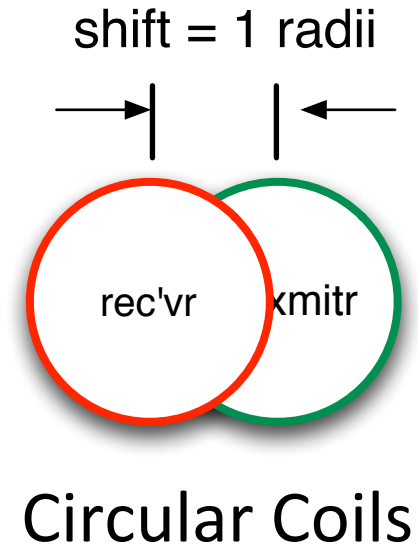
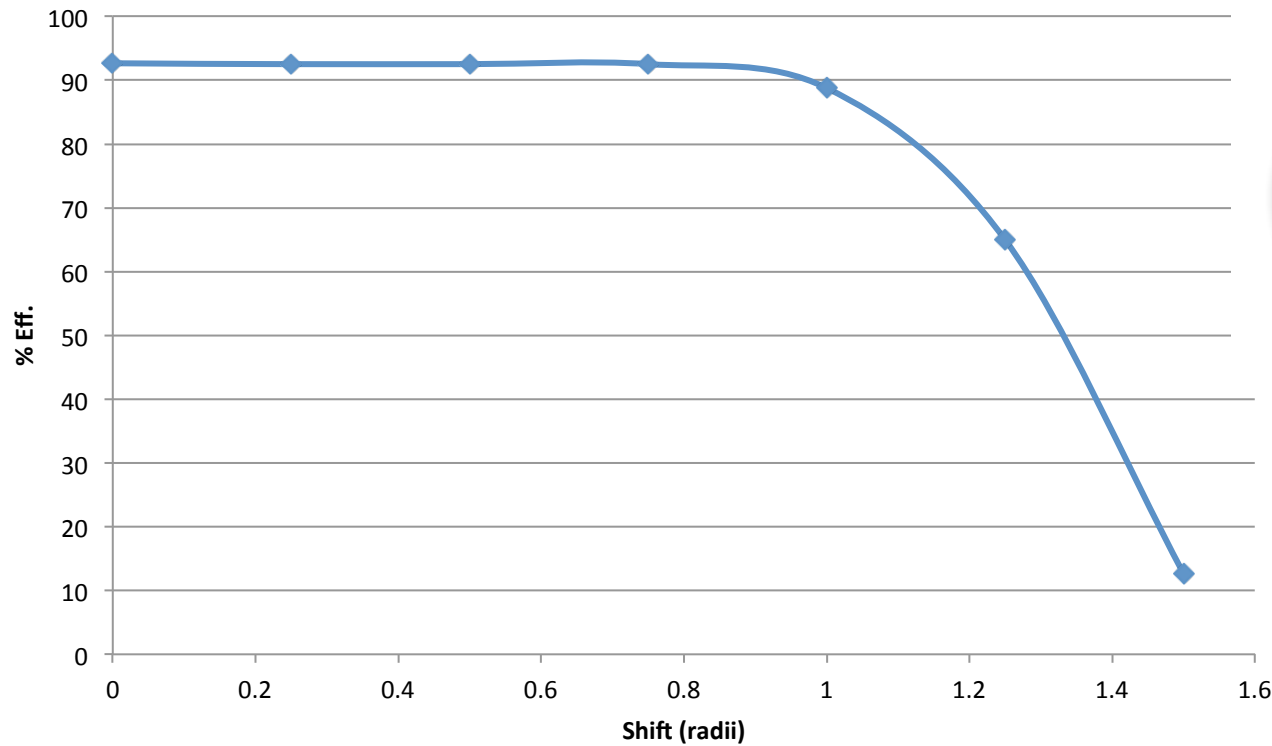
Defilippo, Cooke MIT 2015



Position Sensitivity: Off Axis Shift

Coil Shift: x-axis

Efficiency Vs. Shift @ Resonance

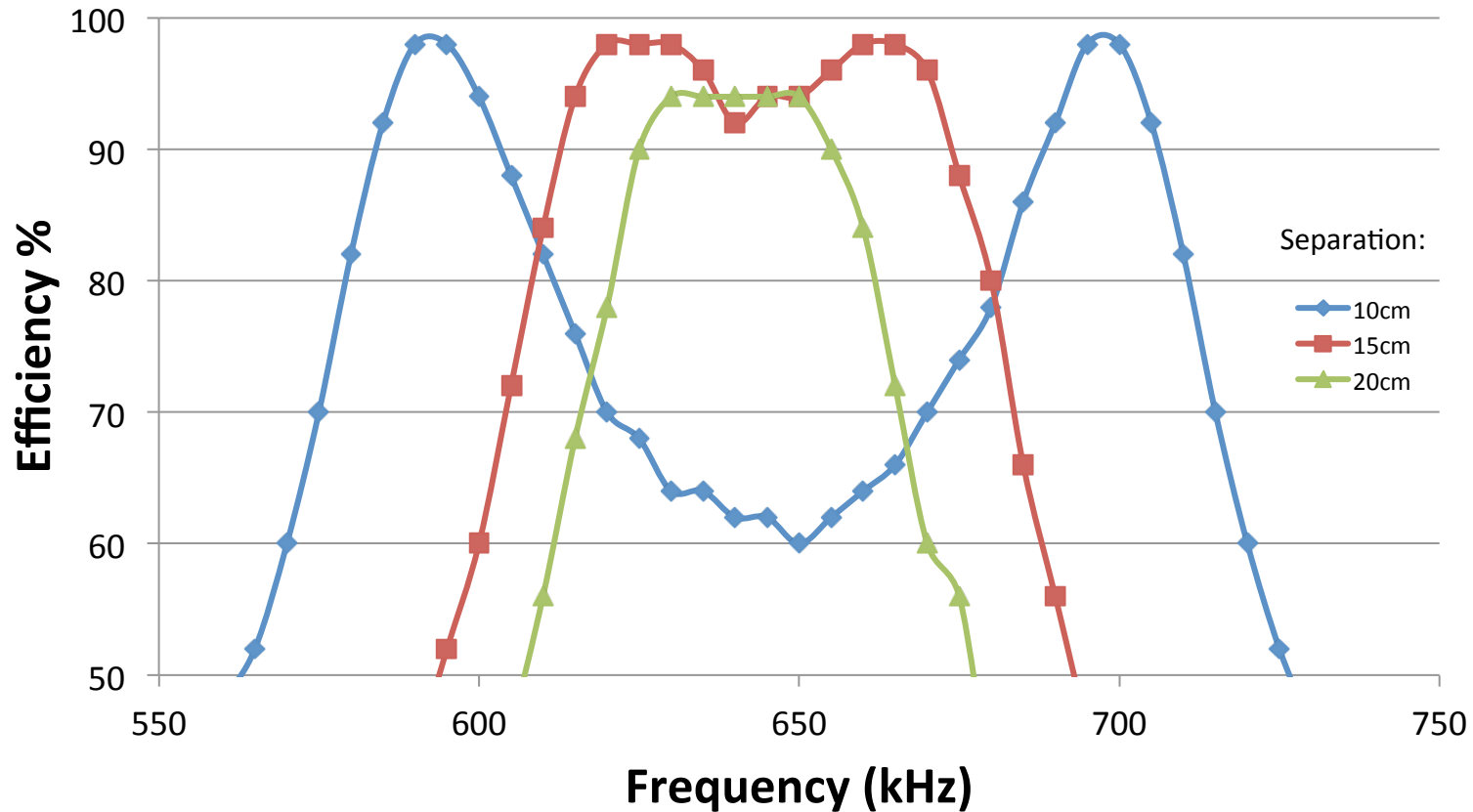


Defilippo, Cooke MIT 2015



Circular Coils in Air, Litz Wire

Measured Frequency Characteristics



Defilippo, Cooke MIT 2015



Lessons from Wireless World

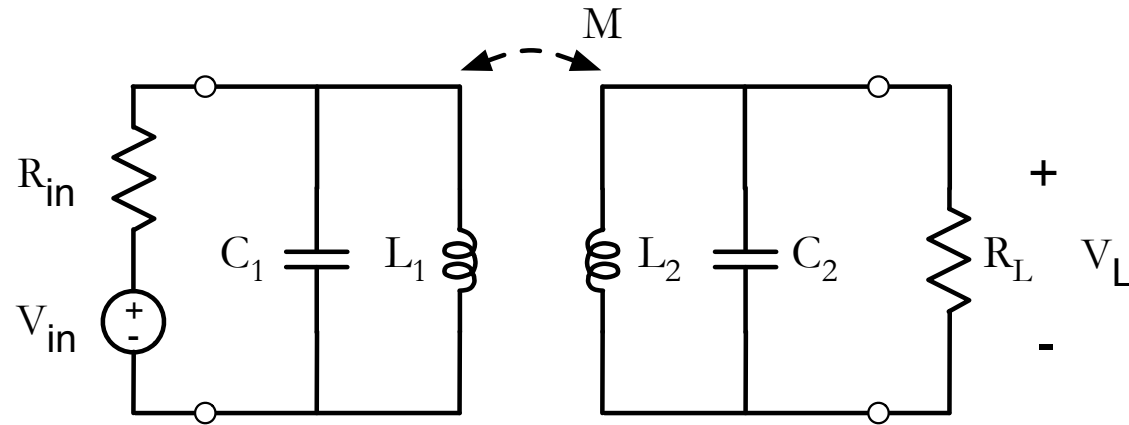
- On-Going Development, Many Applications
 - Electric buses, cars
 - Autonomous-underwater-vehicles (AUV)
 - Medical implants, cell-phones, computers
- Theory and Experimental Agreement
 - High 'Q' is essential for low loss
 - Equivalent circuit models are okay and useful
 - Modal analysis models are okay and useful
 - Efficiency-power FOM = $k [\text{root}(Q_1 Q_2)] \gg 1$
 - Presently data only up to 10s kilowatt level
- Now: **expand concepts to Megawatt Level**

Resonant High-Power Transformers

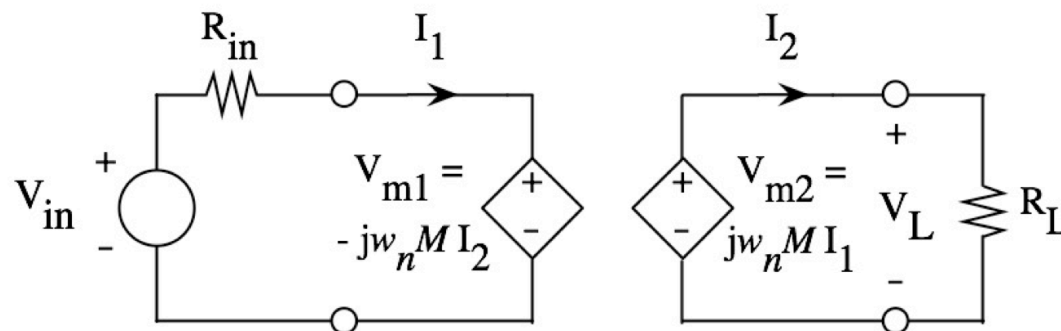
- Maximize power transfer
- Minimize Loss: efficiency constraint
- Apply to 50 – 500 MW, 110 – 500kV designs
 - Requires high frequency
 - Recall transformer size relations
 - To keep similar size:
 - μ reduced by 40,000
 - Frequency increase by $\sim 40,000$ (perhaps $\sim 1\text{MHz}$)
 - No core: less heat, better insulation

Optimized Power Transfer

Equivalent “Loss-Less” Resonant Circuit:



At Resonance:



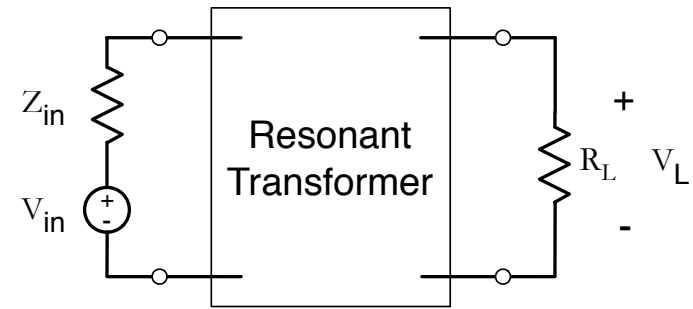
$$k \equiv \frac{M}{\sqrt{L_1 L_2}}$$

Optimized Power Transfer

At Resonance:

Idealized: Neglect Winding Losses

Define Power Transfer Efficiency ($N_1 = N_2$):



$$\text{efficiency} = \eta_2 = \frac{\text{Optimum Power to } R_L \text{ with Transformer}}{\text{Optimum Power to } R_L \text{ without Transformer}}$$

Max Power Transfer:

$$R_L = R_{in} \equiv R_{opt} \quad k \equiv \frac{M}{\sqrt{L_1 L_2}}$$

Opt. Power Design: ($L_1 = L_2$)

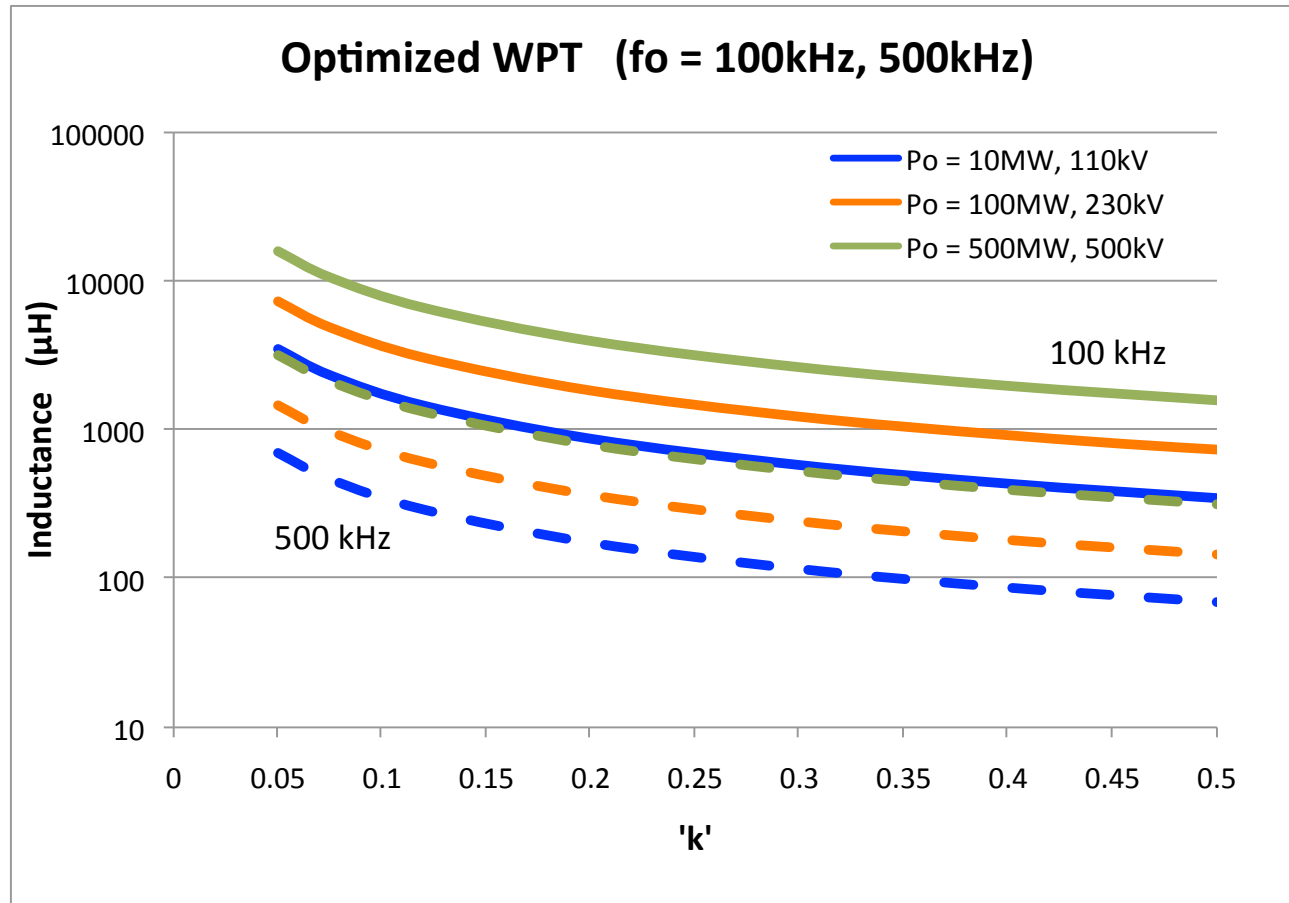
Max Power, fixed ω , fixed M :

$$R_L^2 = \omega_n^2 M^2$$



$$L_{opt} = \frac{(V_L)^2}{2\pi f_o k P_{out}}$$

10 to 500 MW: Core-Less Design



Spiral Coil: 25 turns, 1.5" wire, 0.5" spacing, 24" ID: \rightarrow 1000 μH

Spiral Coil = Shell-Form Pancake Coil



10 to 500 MW: Core-Less Design

- Coils
 - Control & Design separation for optimum 'k'
 - Employ existing coil technology
 - spiral 'pancake' coils
 - Like in shell-form transformers
 - Exhibit better high frequency response
 - Control inductance for low-loss
 - Minimum resistance possible
 - fewer turns, large area wire (Litz)
- Without Core
 - No Core losses
 - More space for cooling
 - More space for insulation

Limitations, Needs

- **Power Electronics** (at the megawatt level !!!)
 - Costs for steel core → power electronics
- Non-magnetic coil supports
 - Magnetic forces still very large
- Stray field control
 - Use ferrite material external to windings
 - Example auto and bus chargers
- High 'Q' coils
 - Low losses, maybe **superconductivity ??**
 - Examples of superconductivity classical transformers

Conclusions

- Classical transformers
 - Costly; large heavy steel cores
 - Complicated construction
- Disruptive approach: eliminate the steel core
 - Use magnetically **coupled resonance**
 - Borrowed concepts from wireless power
 - No steel core
 - High frequency operation, needs power electronics
 - Structure of Pancake coils consistent with wireless power
 - High Q coils essential
- Added Bonus
 - Use power electronics to go 'breaker-less' ???

Thank You

And Thanks to the 'Internet' for Many Great Images

